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**Project: IEEE P802.15 Working Group for Wireless Personal Area Networks (WPANs)**

**Submission Title:** [Basic transmission characteristics of 802.15.4ad SUN OFDM LR payload and frequency hopping method between users]

**Date Submitted:** 13 March 2025

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**Re:** [Wireless Next Generation, Long Range extension enhancements to 802.15.4-2020]

**Abstract:** Show basic transmission characteristics of 802.15.4ad SUN OFDM LR payload and propose frequency hopping method between users. A part of this contribution supported from the commissioned research (No. JPJ012368C05101) by National Institute of Information and Communications Technology (NICT) , Japan is included.

**Purpose:** Show basic transmission characteristics of 802.15.4ad SUN OFDM LR payload and propose frequency hopping method between users.

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# **Basic transmission characteristics of 802.15.4ad SUN OFDM LR payload and frequency hopping method between users**

March 13, 2025

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and Goro Kawabuchi

Kyoto University

# Background

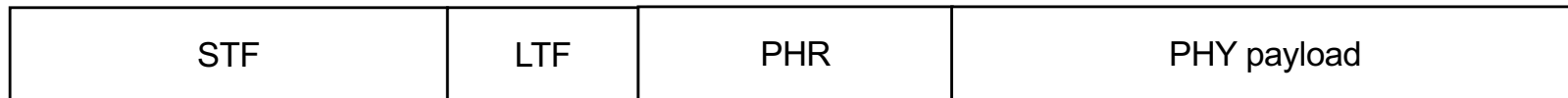
- We have already proposed IEEE 802.15.4 SUN OFDM LR in 15-25-0035r0
- This proposal was an SC-OFDM (Single Carrier OFDM) system that used frequency hopping to first select a frequency and then transmit a block consisting of a pilot signal and information symbols at that selected frequency
- However, the transmission characteristics of this system in an AWGN and multipath fading environment have not yet been evaluated
- In this contribution, we first show the basic transmission characteristics of the SUN OFDM LR3 proposed by authors

# Proposed 802.15.4 SUN-OFDM LR (15-25-0035r0)

|                            |       | Option LR1   | Option LR2 | Option LR3 |
|----------------------------|-------|--|------------|------------|
| Channel spacing            |       | 200 kHz  |            |            |
| Subcarrier spacing         |       | 31.25/3 kHz  |            |            |
| DFT size                   |       | 16   | 16         | 16         |
| Number of subcarriers used |       | 14   | 14         | 14         |
| Num. of data-subcarriers   |       | 12   | 6          | 1          |
| OFDM symbol duration       |       | 120 us   |            |            |
| Guard interval             |       | 24 us  |            |            |
| Primary modulation scheme  |       | BPSK   |            |            |
| Coding Scheme and rate     |       | Convolutional code (Constraint length: 7)<br>Coding rate 1/2 |            |            |
| Spreading factor           |       | 8 (MCS 0), 4 (MCS1), 2(MCS 2)                                |            |            |
| Data rate for PSDU (kb/s)  | MCS 0 | 6.25   |            | 0.521      |
|                            | MCS 1 |  | 6.25       | 1.041      |
|                            | MCS 2 |  |            | 2.083      |

# Frame format SUN-OFDM LR (15-25-0035r0)

OFDM LR1 and LR2 (Same as SUN-OFDM option 4)

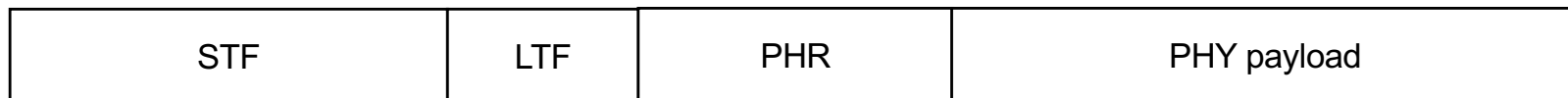


4 OFDM symbols  
(480 us)

2.5 OFDM symbols  
(300 us)

120us/ OFDM symbol

OFDM LR3

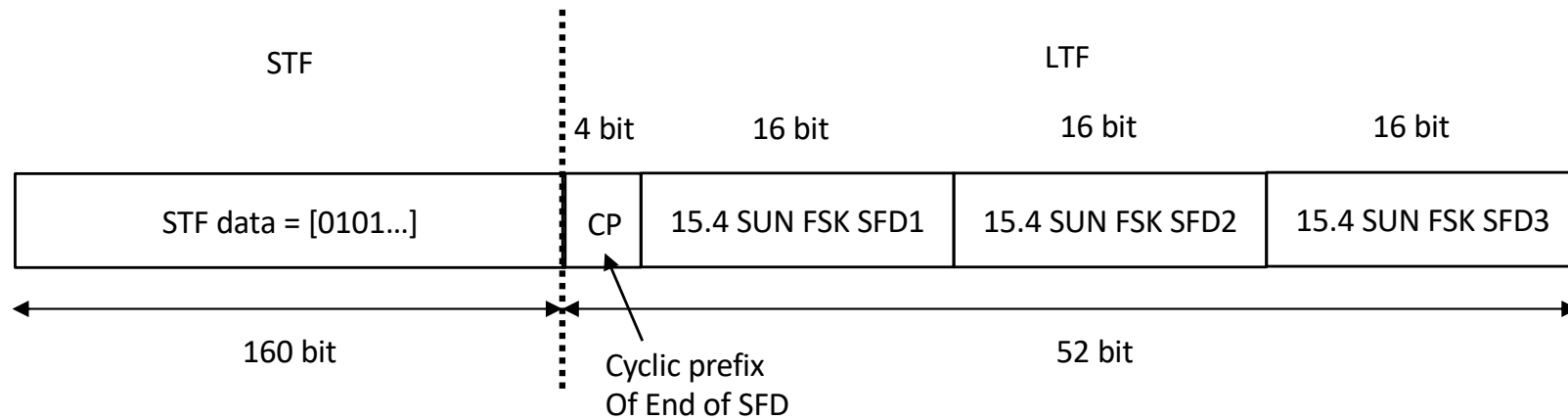


160 bit

52 bit

120us/symbol= 8.33 ksymbol/s

# Proposed SHR of SUN-OFDM LR3 (15-25-0035r0)



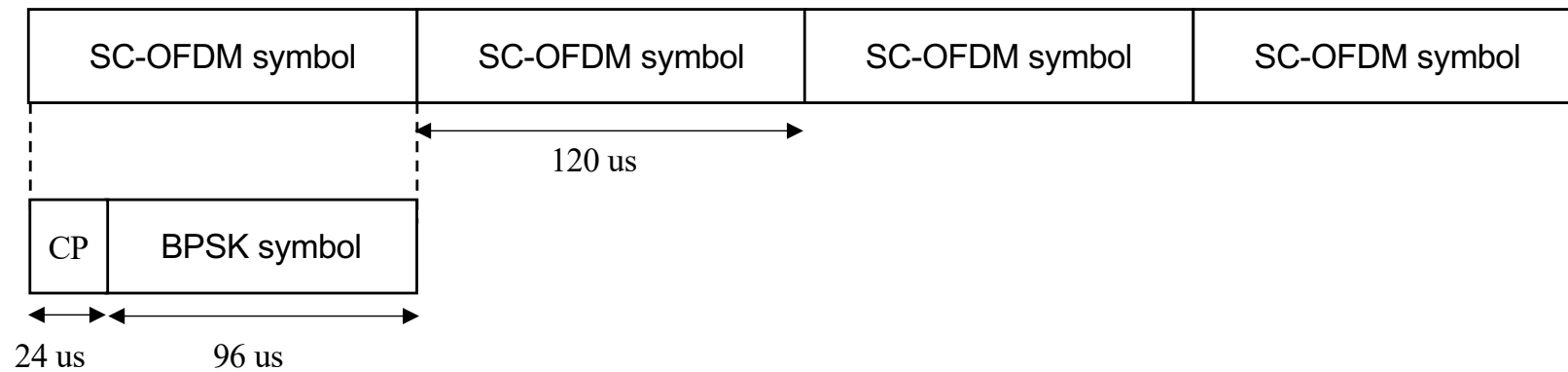
## STF

- Preamble is 160 bits
- Repetitive data of 01 or other sequences are used in 15.4-2024 SUN FSK

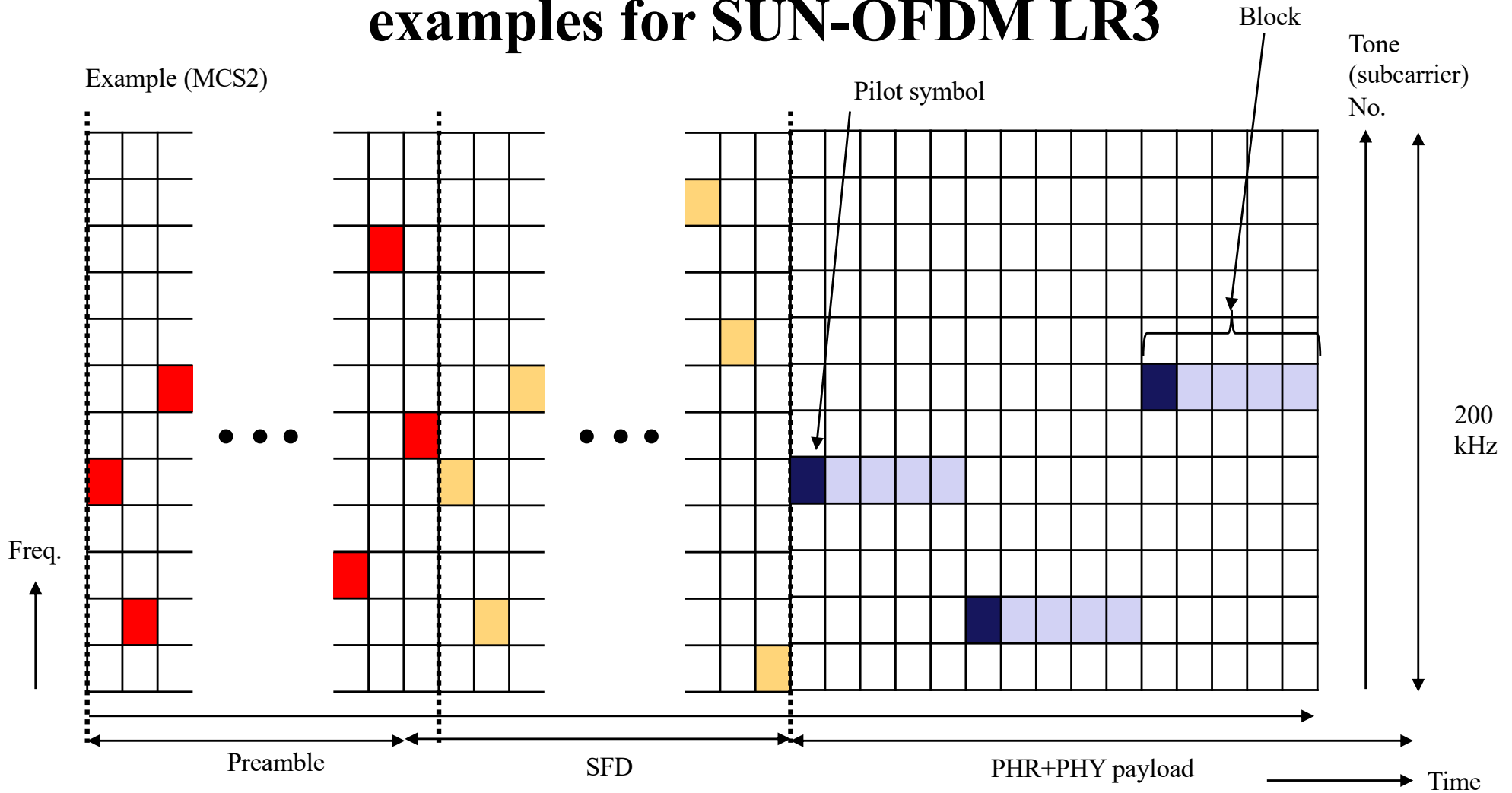
## LTF

- SFD1, SFD2, and SFD3 are the same as those used in 15.4-2024 SUN FSK
- Basically, different ones are used, but some of them may be the same
- Each SFD has already been commercialized and its characteristics are well known
- The last 4 bits of the SFD are added as CP at the beginning

# PHR & PHY payload of SUN-OFDM LR3 (15-25-0035r0)



# Frame construction and frequency hopping examples for SUN-OFDM LR3

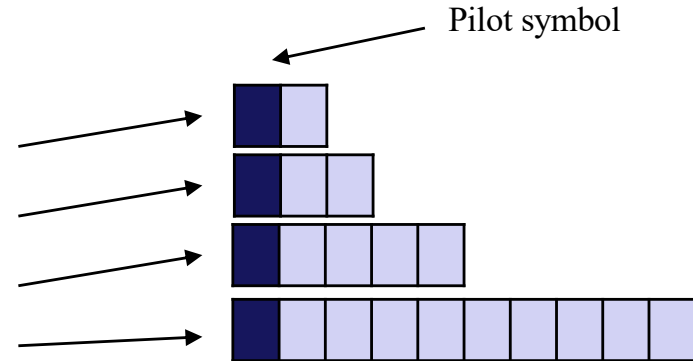


- The hopping pattern should be set considering the coherent bandwidth of the assumed radio propagation channel.
- If each symbol is spread, perform a hop for each symbol that is spread.



# Simulation parameters

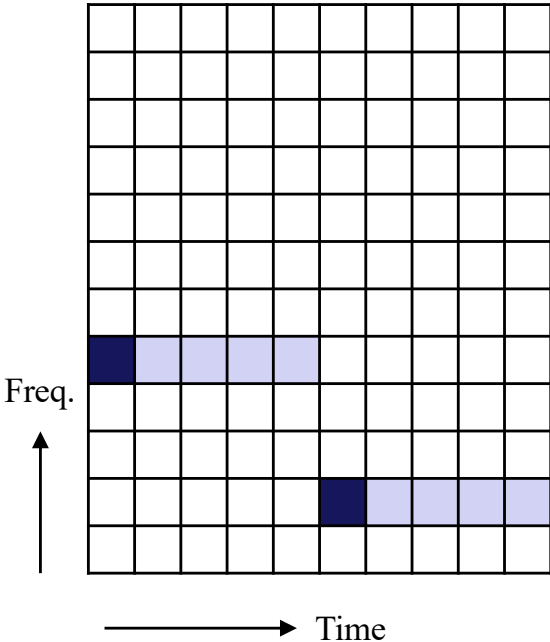
- Block structure
  - BL1: 1 Pilot symbol and 1 Data symbols
  - BL2: 1 Pilot symbol and 2 Data symbols
  - BL4: 1 Pilot symbol and 4 Data symbols
  - BL8: 1 Pilot symbol and 8 Data symbols



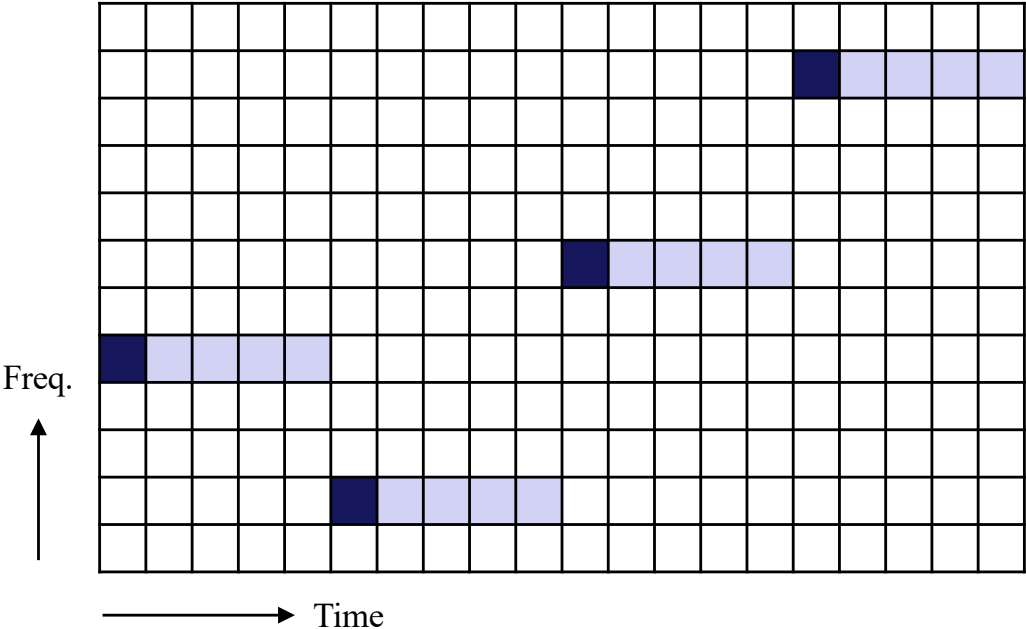
- Of the 14 subcarriers (Option 4), the subcarriers that each user selects as FH(Frequency Hopping)
  - Nsub 2: 2 subcarriers
  - Nsub 4: 4subcarriers
  - Nsub 8: 8 subcarriers
  - Nsub14: 14 subcarriers
- In this simulation, the subcarriers selected as FH are randomly selected from among 14 subcarriers.
- Each block does not overlap on the time axis.
- After estimating the propagation characteristics using the pilot symbol at the beginning of each block, the data in each block is demodulated.

# Example of frame construction

BL=4, Nsub=2



BL=4, Nsub=4



## Simulation parameters (others)

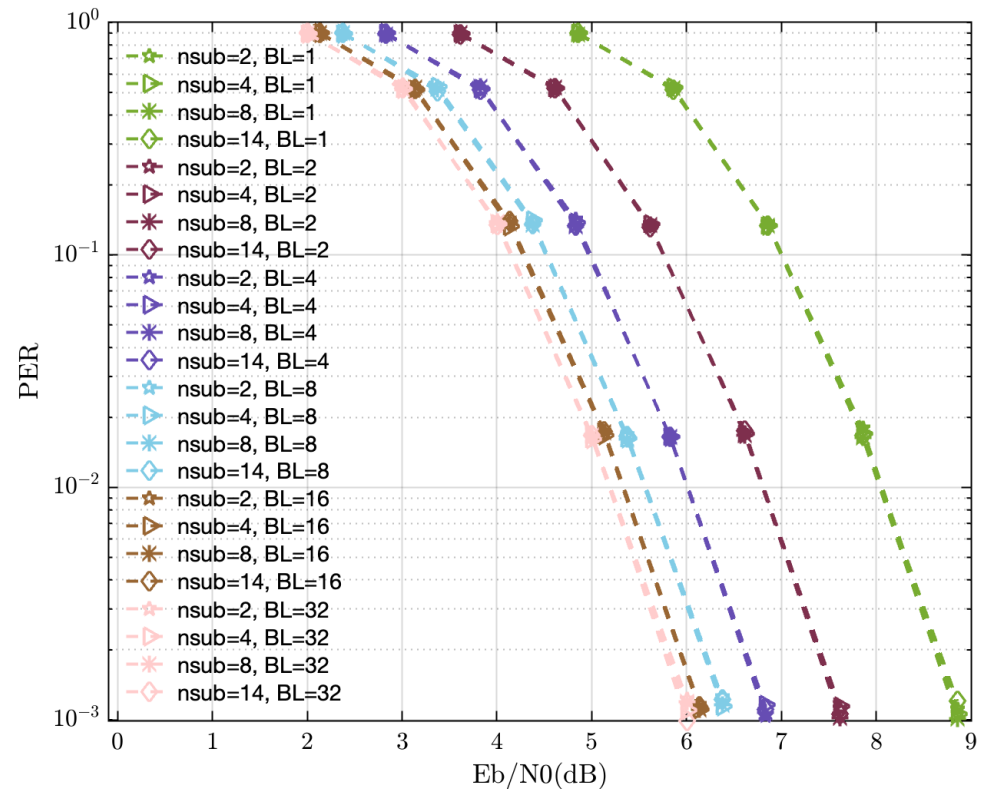
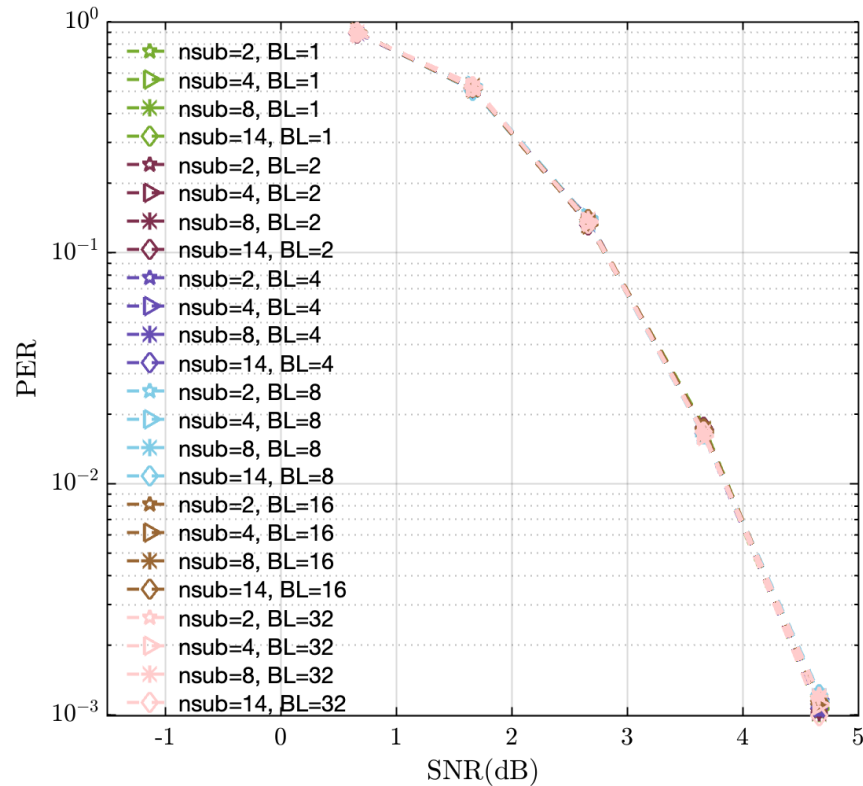
- Spreading: No spreading
- Channel model: GSM typical urban, IEEE 802.22 Profile A
- Doppler frequency: 0.6 Hz and 6 Hz
- Packet size: 20 byte
- Interleave size: Random
- Evaluation index: Packet Error Rate (PER)
- Padding bit: 2bit
- Tail bit: 6bit
- Guard interval: 1/4

# SNR and received signal power conversion

- In the case of Noise Figure (NF) =0dB and bandwidth =31.25/3 kHz, noise power should be -133.8 dB

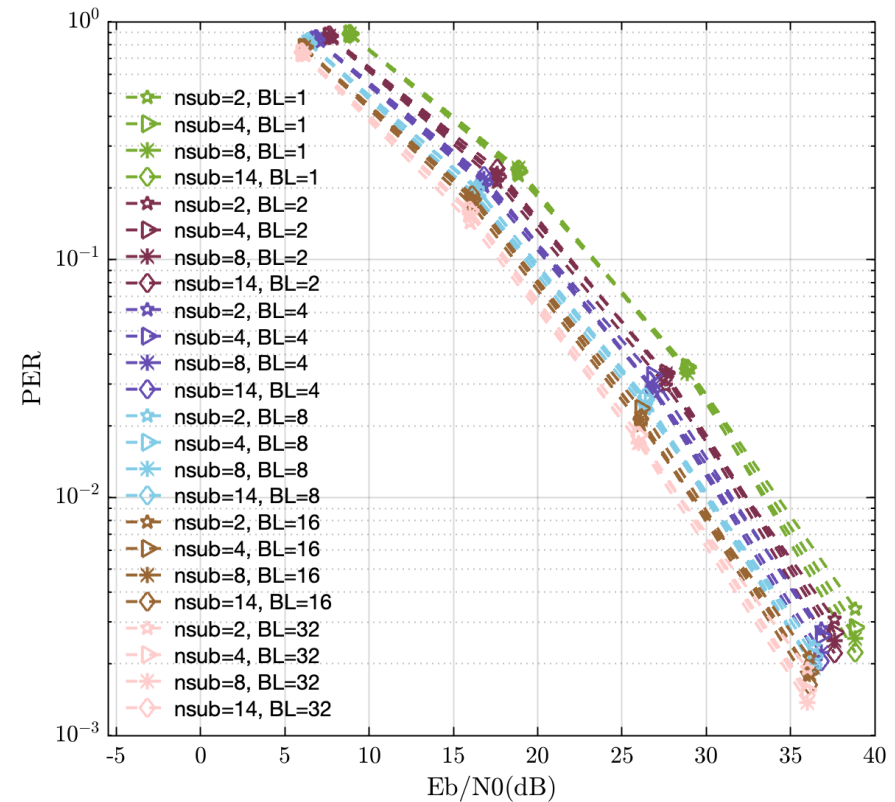
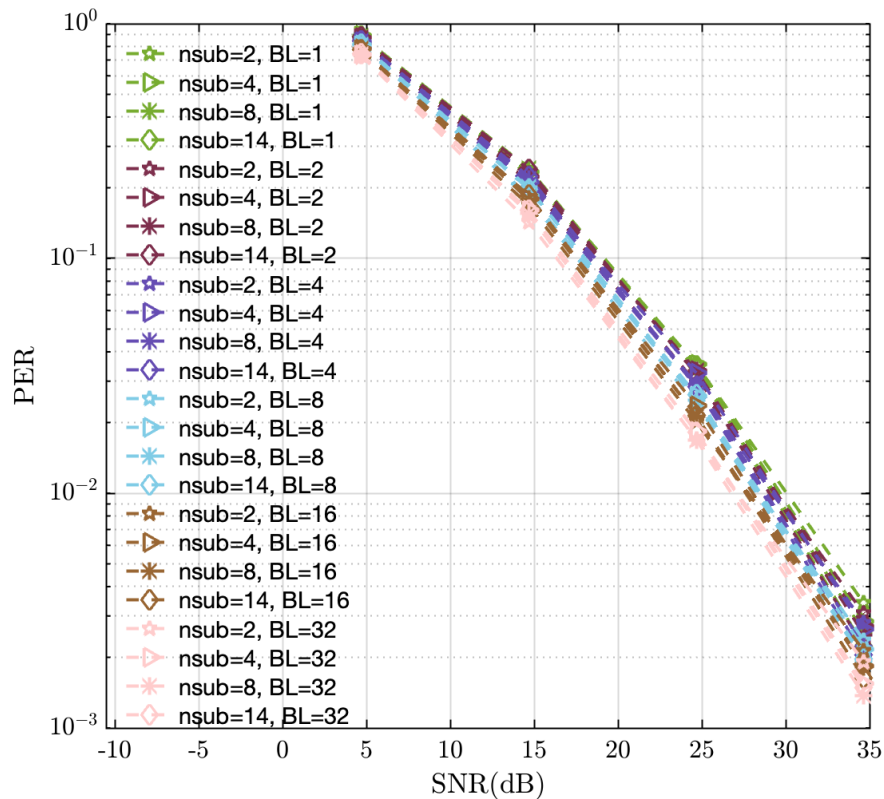
| SNR(dB) | Received power (dBm) |
|---------|----------------------|
| -10     | -143.8 dBm           |
| -5      | -138.8 dBm           |
| 0       | -133.8 dBm           |
| 5       | -128.8 dBm           |
| 10      | -123.3 dBm           |

# PER under AWGN



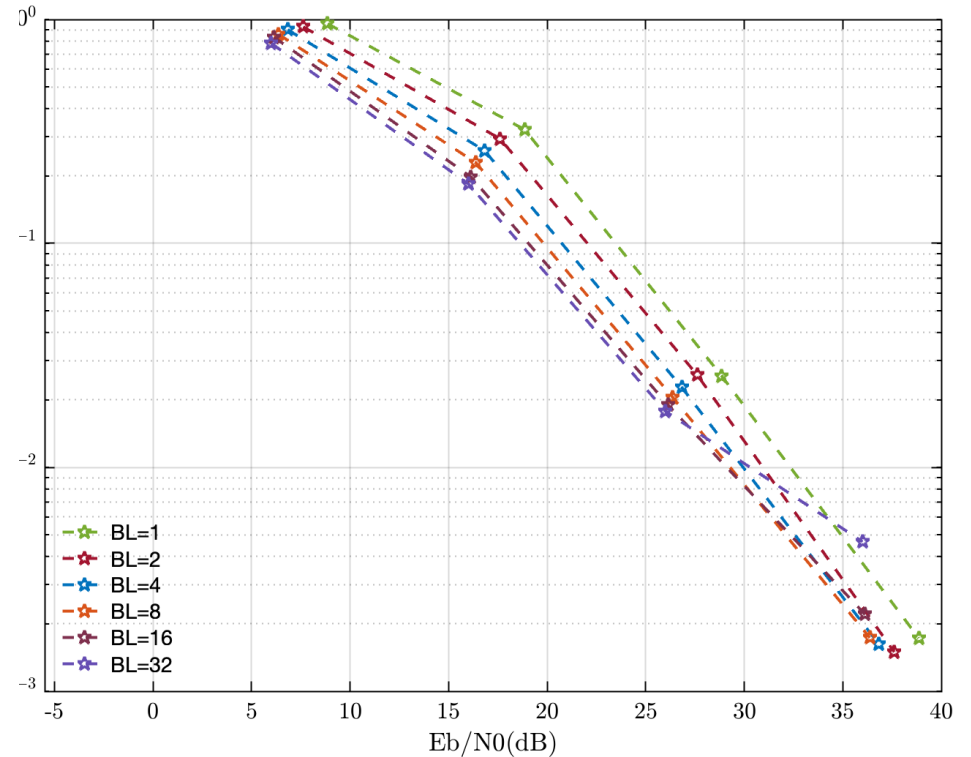
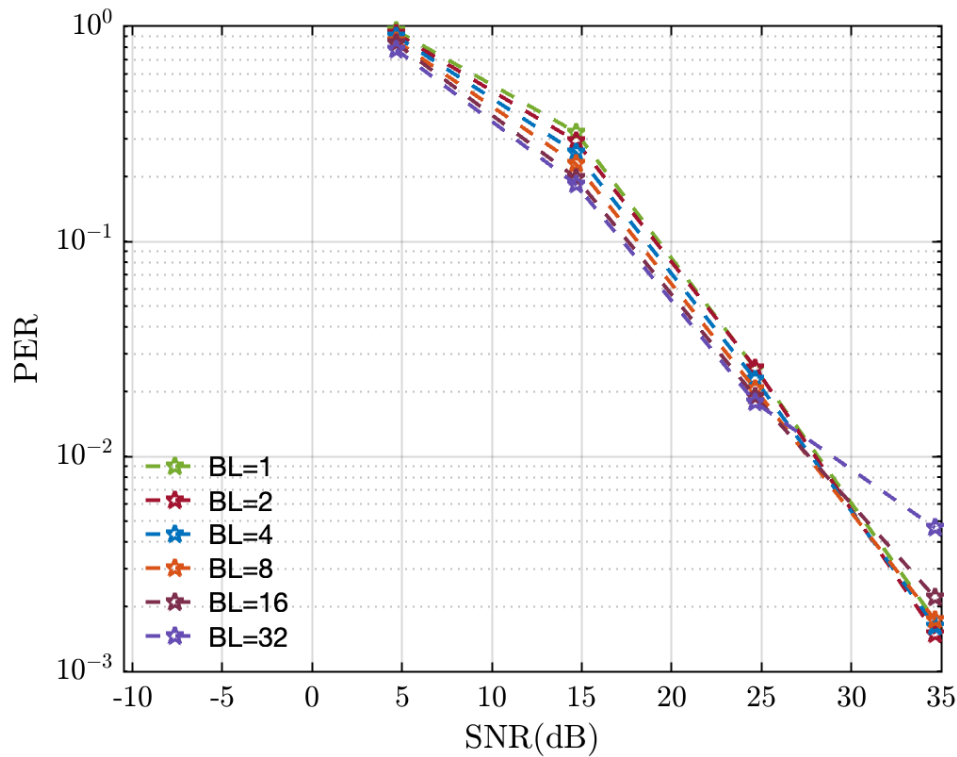
- The required SNR to achieve  $PER=10^{-2}$  is around 3.8dB (Received power:  $-130$  dBm).
- As the BL becomes smaller, efficiency decreases and the required  $E_b/N_0$  increases.

# PER under GSM Typical Urban (fd=0.6Hz)



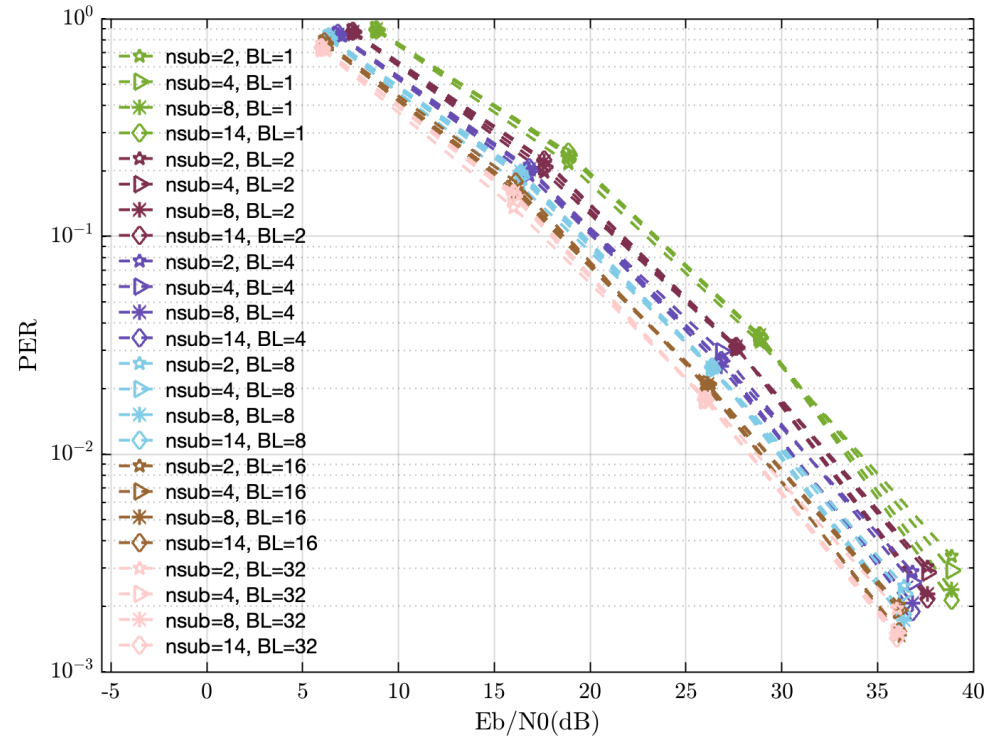
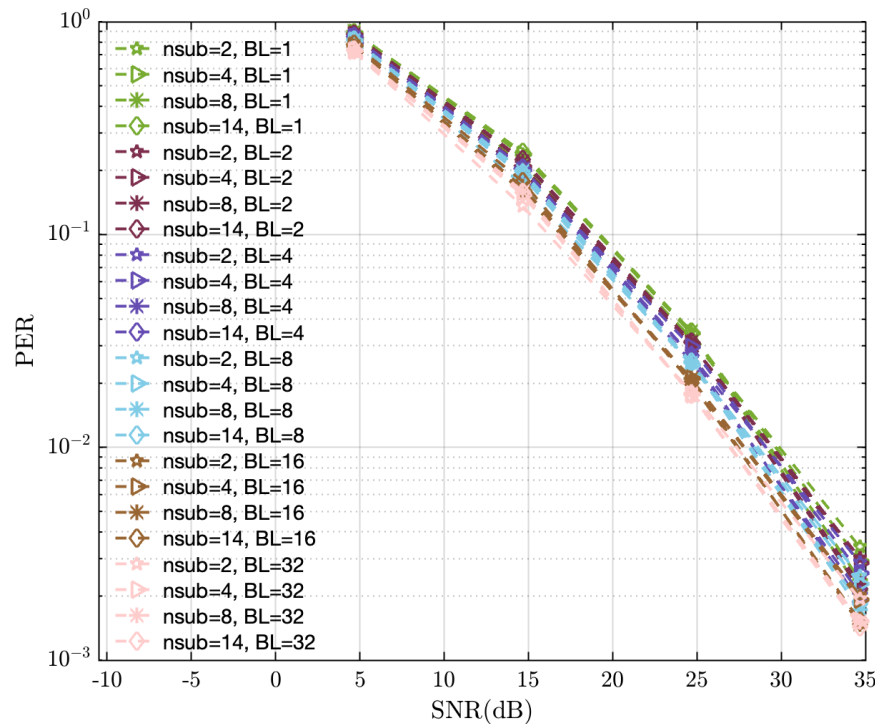
- The required SNR to achieve PER=10<sup>-2</sup> is around 28 dB (Received power: -105.8dBm).
- Because pilot signals are inserted in each block, the shorter the BL, the longer the time it takes to transmit a single packet. As a result, it becomes more susceptible to fading.

# PER under GSM Typical Urban ( $f_d=6\text{Hz}$ , $N_{\text{sub}}=14$ )



- The required SNR to achieve  $\text{PER}=10^{-2}$  is around 27.5 dB (Received power:  $-106.3$  dBm).
- The longer the BL, the less influence the pilot signal estimation has on the entire BL and the worse the PER characteristics.
- BL should be 16 or below.

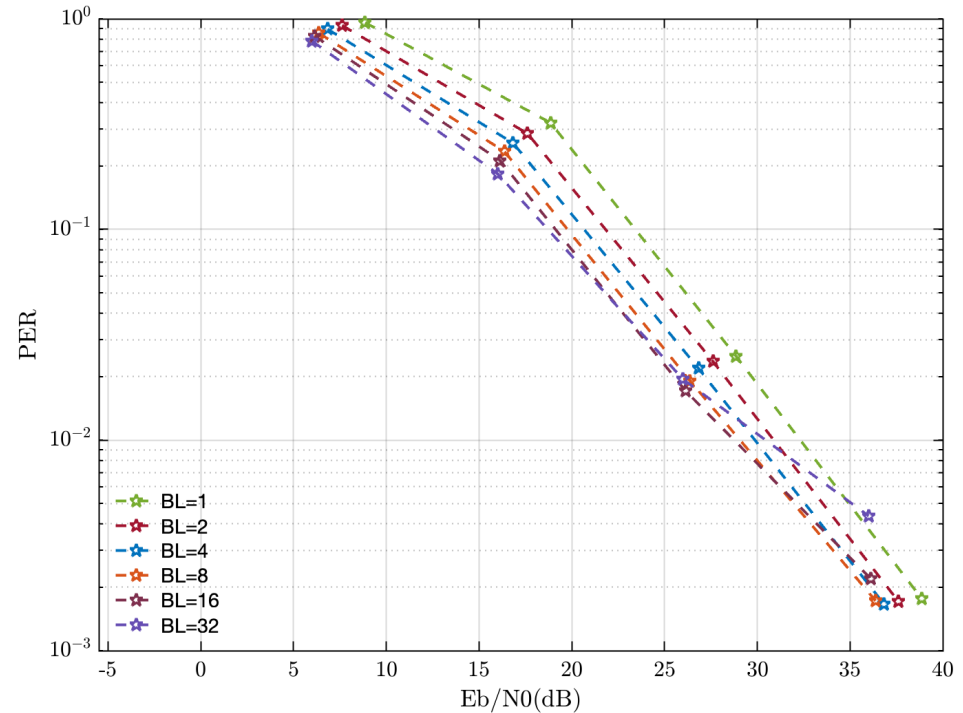
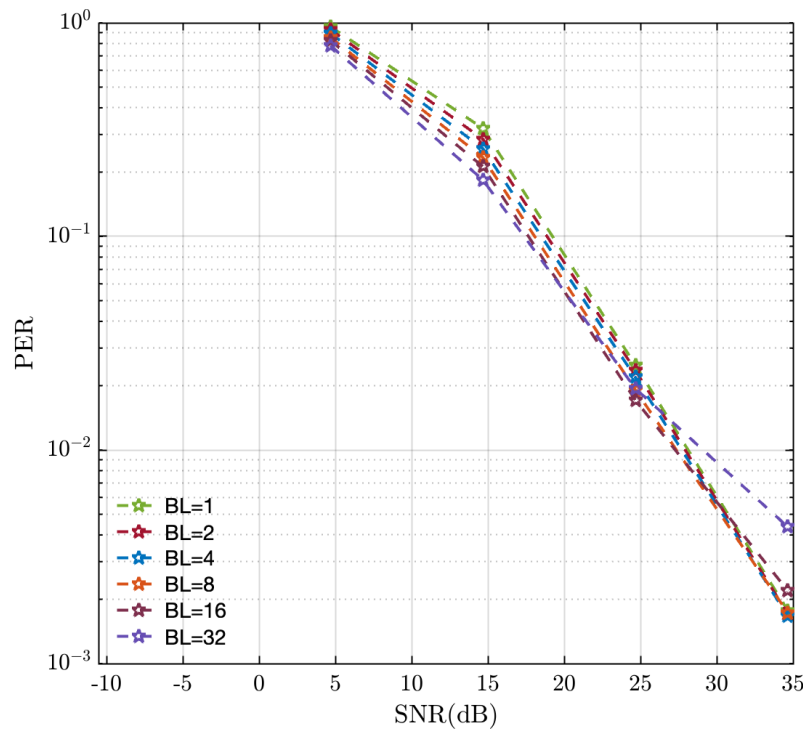
# PER under IEEE 802.22 Profile A ( $f_d=0.6\text{Hz}$ )



- The required SNR to achieve  $PER=10^{-2}$  is around 27.5 dB (Received power:  $-106.3$  dBm).
- This channel model has strong frequency selectivity because it receives long delay waves.
- However, the guard interval is longer than the maximum delay time, so there is no significant impact on PER.
- The longer the BL, the less influence the pilot signal estimation has on the entire BL and the worse the PER characteristics.



# PER under IEEE 802.22 Profile A ( $f_d=6\text{Hz}$ , $N_{\text{sub}}=14$ )



- The required SNR to achieve  $\text{PER}=10^{-2}$  is around 27.5 dB (Received power:  $-106.3$  dBm).
- This channel model has strong frequency selectivity because it receives long delay waves.
- The longer the BL, the less influence the pilot signal estimation has on the entire BL and the worse the PER characteristics.
- BL should be 16 or below.

# Mapping scheme

- There are various methods of block mapping based on  $N_{sub}$ .
  - Random mapping : One subcarrier is selected from among the available subcarriers at each time.
  - Non-Random mapping: Mapping is performed based on a mapping scheme.

# A proposal of mapping scheme (Nsub=14)

The tone(subcarrier) number assigned to each blocks:  $-\frac{N_{FFT}}{2}, -\frac{N_{FFT}}{2} + 1, \dots, \frac{N_{FFT}}{2} - 1$  ( $i = 1, 2, 3, \dots$ )

$$n_{sub,i} = \text{mod}\left(\left\{\frac{N_{FFT}}{2} \times \text{mod}(i - 1, 2) + \frac{N_{FFT}}{4} \times \text{mod}\left(\text{floor}\left(\frac{i - 1}{2}\right), 4\right) + \text{mod}\left(\text{floor}\left(\frac{i - 1}{4}\right), \frac{N_{FFT}}{4}\right)\right\}, N_{FFT}\right) + 1 - \frac{N_{FFT}}{2}$$

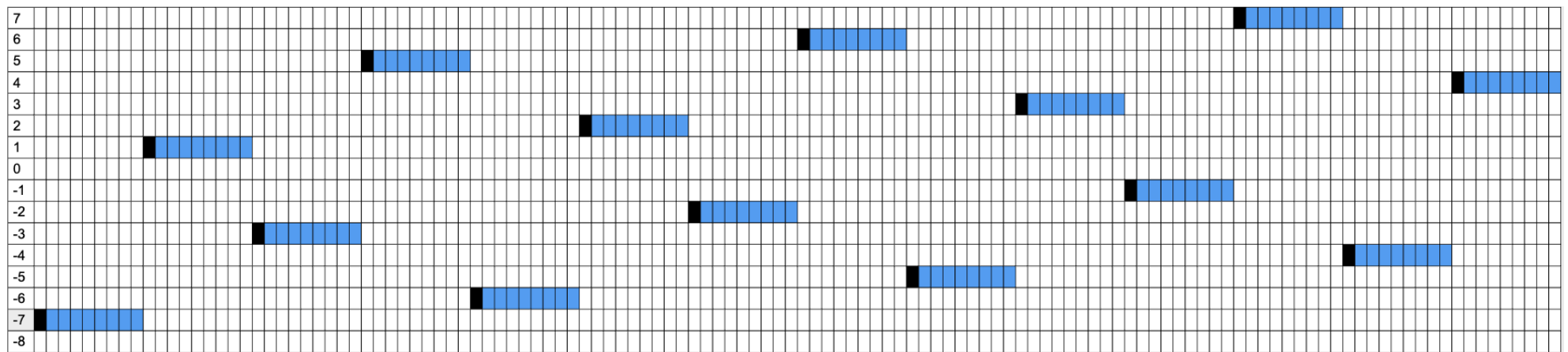
Inactive subcarriers are excluded.

In the case of Option4 ( $N_{FFT}=16$ ), the subcarriers selected in the  $i$ -th block are as follows.

$$n_{sub,i} = -7, 1, -3, 5, -6, 2, -2, 6, -5, 3, -1, 7, -4, 4, \dots$$

Assigned  
subcarrier<sup>-7</sup>

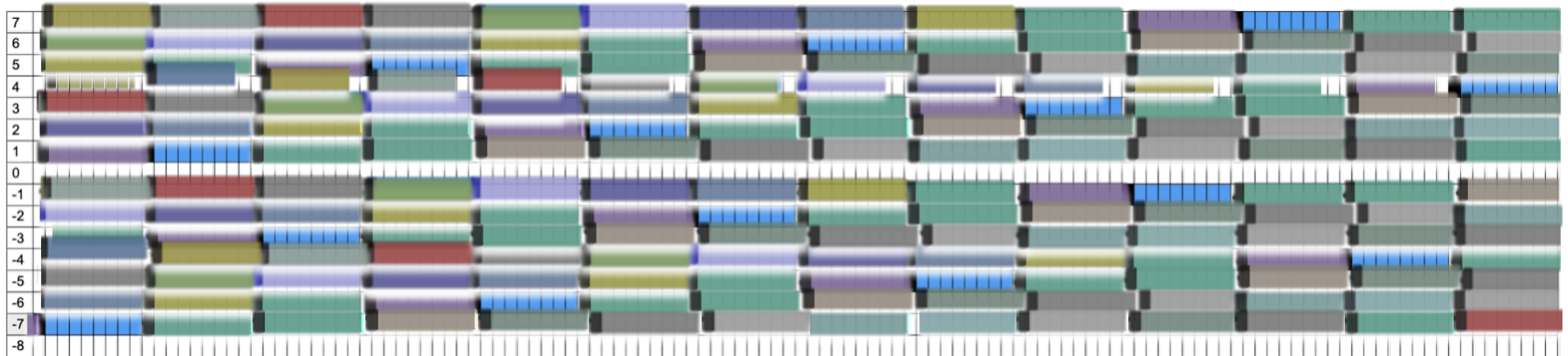
1      -3      5      -6      2      -2      6      -5      3      -1      7      -4      4



BL=8

# Mapping scheme

- By arranging the subcarriers in a cyclical pattern, multiple transmitters can use the bandwidth efficiently (without overlap)
- The same number of transmitters as the number of active subcarriers can be accommodated in the same bandwidth with minimal interference
- For example, one possible method would be to assign a cyclic shift of one subcarrier on the frequency axis to the subcarriers allocated and used in Slide 19, and then assign the pattern to other user.



# Pilot Generator

Use the existing pilot generation method as is

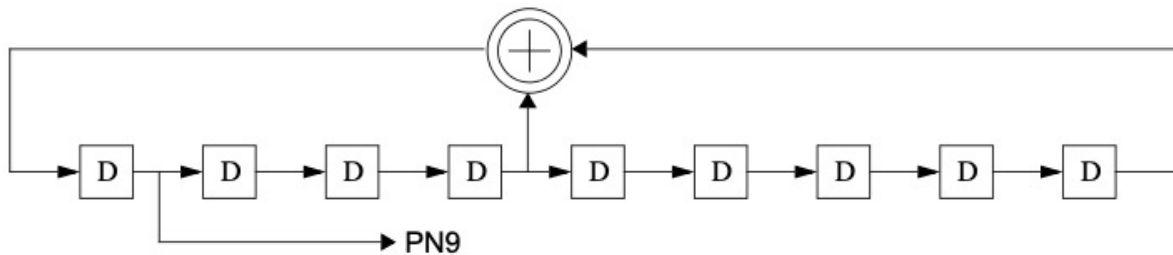


Figure 16-2—Schematic of the PN generator

The seed in the PN9 shall be all ones: “1 1 1 1 1 1 1 1.” The PN9 shall be reinitialized to the seed after each packet (either transmit or receive).

The PN9 generator is clocked using the seed as the starting point and enabled after the first clock cycle. For example, the first 30 bits out of the PN9, once it is enabled, would be as follows:

$$PN9_n = 0_0 0_1 0_2 0_3 1_4 1_5 1_6 1_7 0_8 1_9 1_{10} 1_{11} 0_{12} 0_{13} 0_{14} 0_{15} 1_{16} 0_{17} 1_{18} 1_{19} 0_{20} 0_{21} 1_{22} 1_{23} 0_{24} \\ 1_{25} 1_{26} 0_{27} 1_{28} 1_{29}$$

Ref.: IEEE Std 802.15.4-2020, p.524

Example:  $p_1, p_2, p_3, p_4, p_5, p_6, \dots = -1, -1, -1, -1, 1, 1, \dots$

# Spreading scheme (Payload)

$$d_k = \begin{cases} 1 \\ -1 \end{cases} (k=1,2,3,4,5,6,\dots) \quad \Rightarrow \quad e_k (k=1,2,3,4,5,6,\dots)$$

Spreading factor:2

Data:4(input)×2(Spreading)  
Pilot:2

$$e_{2k-1} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{Data \ Tones}/2) - 1)}{4}\right)$$

$$e_{2k} = d_k$$

Spreading factor:4

Data:2×4  
Pilot:4

$$e_{4k-3} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{4k-2} = d_k \exp\left(j \frac{2\pi(3 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{4k-1} = d_k$$

$$e_{4k} = d_k \exp\left(j \frac{2\pi(\text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

| Option | $N_{Data \ Tones}$ |
|--------|--------------------|
| 1      | 12                 |
| 2      | 24                 |
| 3      | 48                 |
| 4      | 96                 |

Spreading factor:8

Data:1×8  
Pilot:8

$$e_{8k-7} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4} + j\pi\right)$$

$$e_{8k-6} = d_k \exp\left(j \frac{2\pi(3 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4} + j\pi\right)$$

$$e_{8k-5} = d_k \exp(j\pi)$$

$$e_{8k-4} = d_k \exp\left(j \frac{2\pi(\text{mod}(k, N_{Data \ Tones}/4) - 1)}{4} + j\pi\right)$$

$$e_{8k-3} = d_k \exp\left(\frac{j2\pi(2 * \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{8k-2} = d_k \exp\left(\frac{j2\pi(3 * \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{8k-1} = d_k$$

$$e_{8k} = d_k \exp\left(\frac{j2\pi(\text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

# Spreading scheme (Payload)

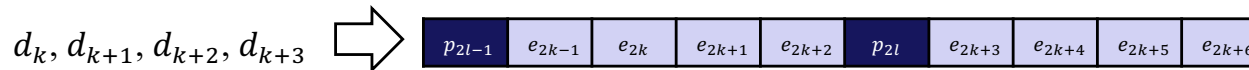
$$d_k = \begin{cases} 1 \\ -1 \end{cases} (k=1,2,3,4,5,6,\dots) \implies e_k (k=1,2,3,4,5,6,\dots)$$

Spreading factor:2

Data:4(input)×2(SF)  
Pilot:2

$$e_{2k-1} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{Data \ Tones}/2) - 1)}{4}\right)$$

$$e_{2k} = d_k$$



| option | $N_{Data \ Tones}$ |
|--------|--------------------|
| 1      | 12                 |
| 2      | 24                 |
| 3      | 48                 |
| 4      | 96                 |

Spreading factor:4

Data:2×4  
Pilot:4

$$e_{4k-3} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{4k-2} = d_k \exp\left(j \frac{2\pi(3 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{4k-1} = d_k$$

$$e_{4k} = d_k \exp\left(j \frac{2\pi(\text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$



Spreading factor:8

Data:1×8  
Pilot:8

$$e_{8k-7} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4} + j\pi\right)$$

$$e_{8k-6} = d_k \exp\left(j \frac{2\pi(3 \times \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4} + j\pi\right)$$

$$e_{8k-5} = d_k \exp(j\pi)$$

$$e_{8k-4} = d_k \exp\left(j \frac{2\pi(\text{mod}(k, N_{Data \ Tones}/4) - 1)}{4} + j\pi\right)$$

$$e_{8k-3} = d_k \exp\left(\frac{j2\pi(2 * \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{8k-2} = d_k \exp\left(\frac{j2\pi(3 * \text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$

$$e_{8k-1} = d_k$$

$$e_{8k} = d_k \exp\left(\frac{j2\pi(\text{mod}(k, N_{Data \ Tones}/4) - 1)}{4}\right)$$



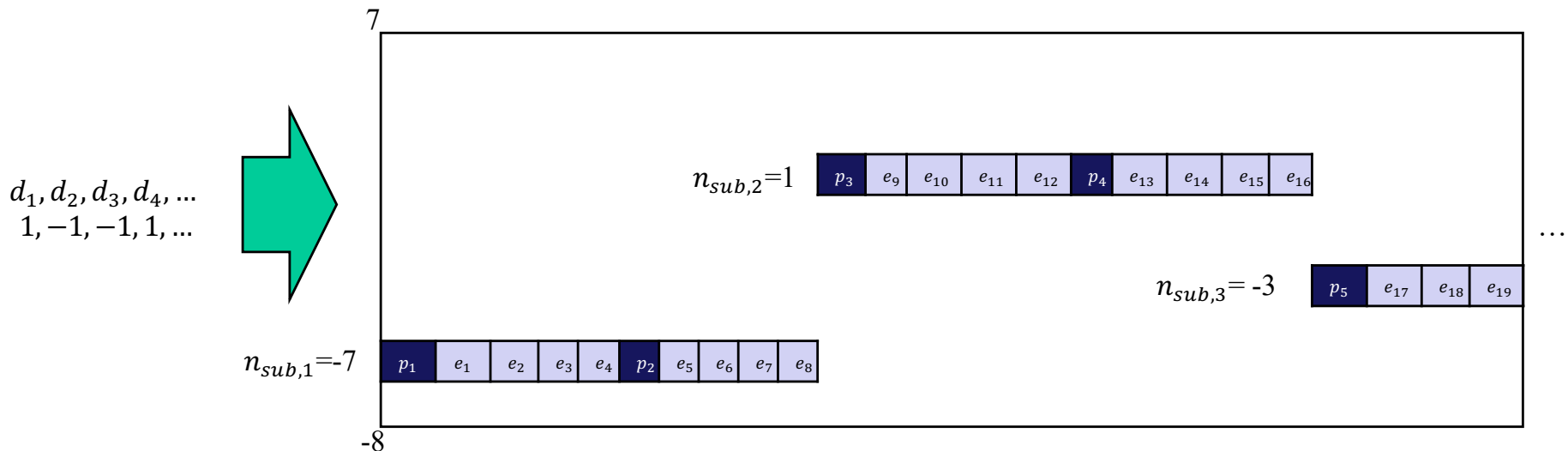
# Example of data spreading (Option=4, SF=2)

Input  $d_1, d_2, d_3, d_4, \dots = 1, -1, -1, 1, \dots$  とすると

Spreading  $e_1, e_2, e_3, e_4, e_5, e_6, e_7, e_8, \dots = 1j, 1, 1j, -1, -1j, -1, -1j, 1, \dots$   $\left\{ \begin{array}{l} e_{2k-1} = d_k \exp\left(j \frac{2\pi(2 \times \text{mod}(k, N_{\text{Data Tones}}/2) - 1)}{4}\right) \\ e_{2k} = d_k \end{array} \right\}$

Pilot  $p_1, p_2, p_3, p_4, p_5, p_6, \dots = -1, -1, -1, -1, 1, 1, \dots$   $\{ \text{PN}_{9N} \}$

Subcarrier  $n_{sub,i} = -7, 1, -3, \dots$

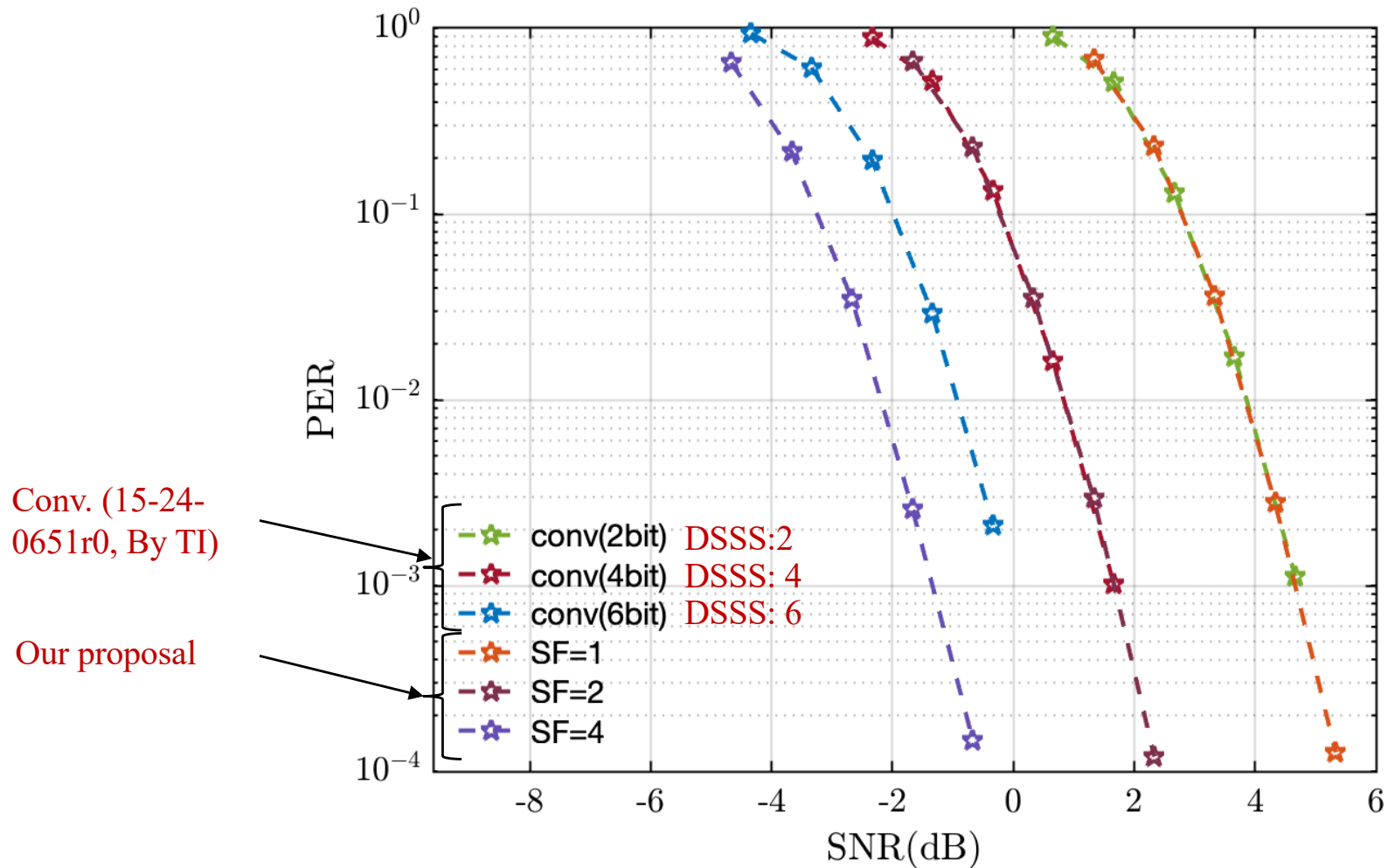




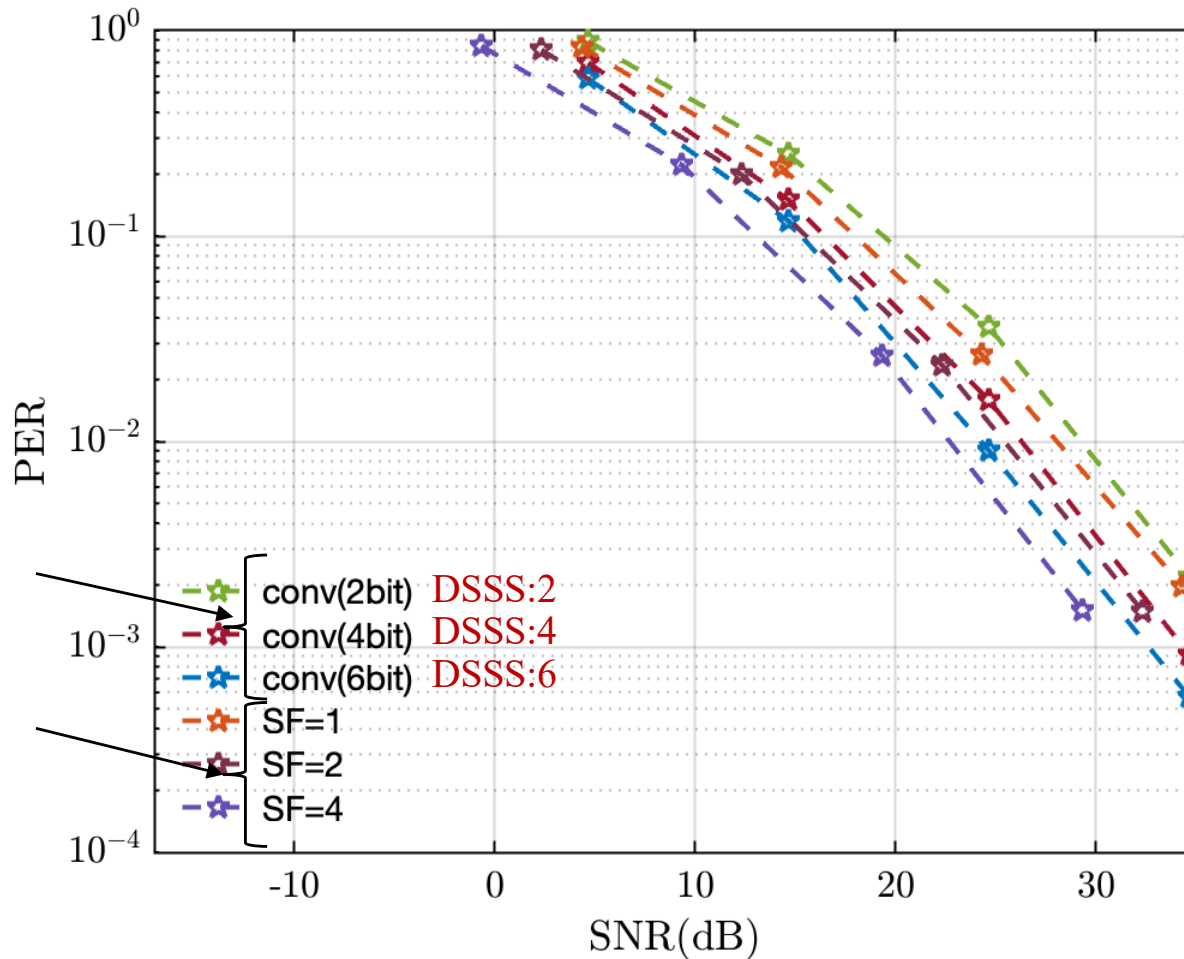
## Simulation parameters (others)

- Spreading factor: 1, 2, 4 (Our proposal) and 2, 4, 6 as Conv. (15-24-0651r0, By TI)
- Channel model: GSM typical urban, IEEE 802.22 Profile A
- Doppler frequency: 0.6 Hz
- Packet size: 20 byte
- Interleave size: Random
- Evaluation index: Packet Error Rate (PER)
- Padding bit: 2bit
- Tailbit: 6bit
- Gurad interval: 1/4

# PER under AWGN ( $f_d=0.6\text{Hz}$ , $N_{\text{sub}}=14$ , $\text{BL}=8$ )



# PER under GSM Typical Urban ( $f_d=0.6\text{Hz}$ , $N_{\text{sub}}=14$ , $BL=8$ )



Conv. (15-24-0651r0, By TI)

Our proposal

# PER under IEEE 802.22 Profile A ( $f_d=0.6\text{Hz}$ , $N_{\text{sub}}=14$ , $BL=8$ )

