

**Project: IEEE P802.15 Working Group for Wireless Personal Area Networks**  
**(WPANs)**

**Submission Title:** THz Wireless Communications: New Opportunities and Challenges

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**Re:** n/a

**Abstract:** This talk focuses on chip-to-chip interconnects where wire-based interconnects are becoming a bottleneck for performance and scalability. Among wire-replacement candidates, wireless interconnect is especially promising because wireless links between chips would circumvent the pin-count problem. Currently investigated mm-wave wireless interconnects face two problems: not enough bandwidth and antennas that are too big for successful integration. Both problems call for terahertz (THz)-range communications. We will talk about THz propagation and channel modeling in chip-to-chip environments.

**Purpose:** Information of IEEE 802.15 IG THz

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# THz Wireless Communications: New Opportunities and Challenges

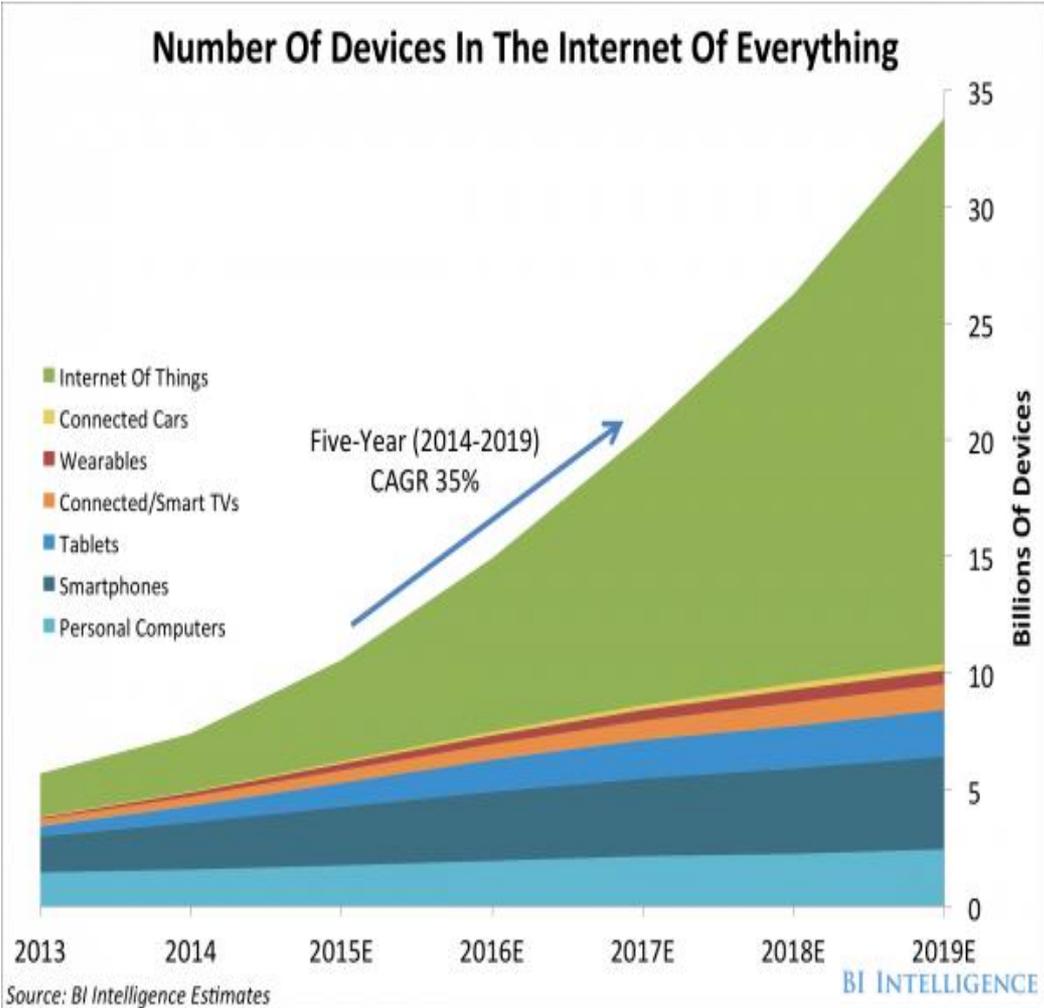
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Prof. Alenka Zajić



January 2017

# ❖ Internet of Everything



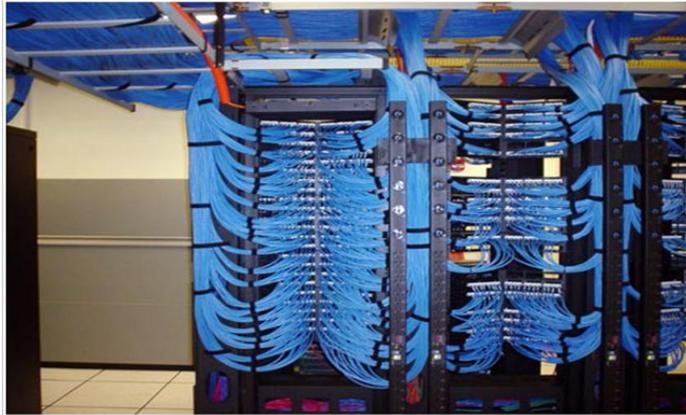
# ❖ Challenges for IoT and Wearable Devices

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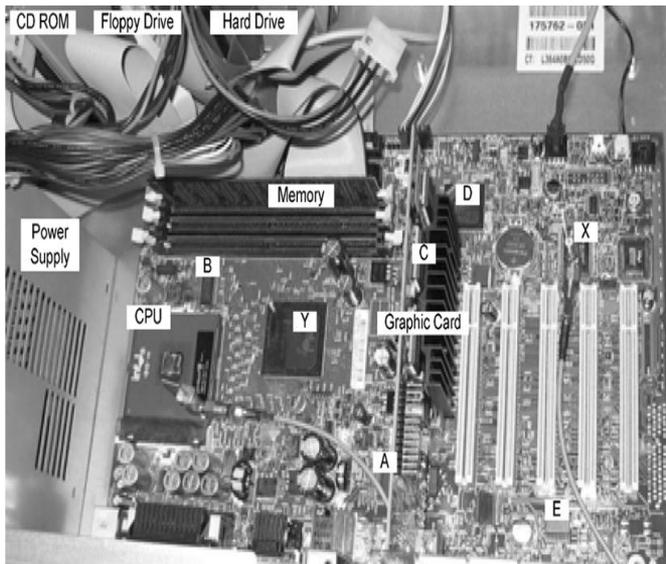
- **Sensing a complex environment** -Innovative ways to sense and deliver information from the physical world to the cloud
- **Connectivity-** Variety of wireless networks needed
- **Cloud is important** - IoT will require significant increase in data storage  needs better rack-to-rack, device-to-device, and chip-to-chip communication
- **Security is vital** - Detecting and blocking malicious activity
- **IoT is complex** IoT application development needs to be easy for all developers, not just to experts
- **Power is critical**

# ❖ Applications That Need THz Communication

## Interconnects in Data Centers



## Chip-to-Chip On Motherboard



## Wearable Devices



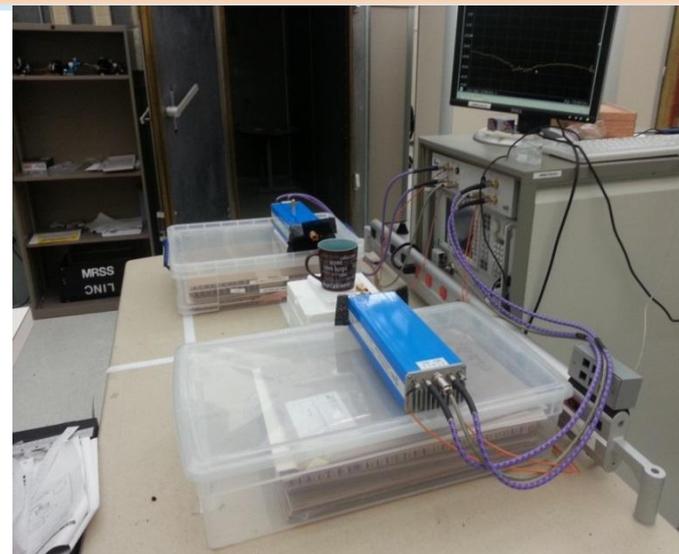
# ❖ Current Wireless Interconnects

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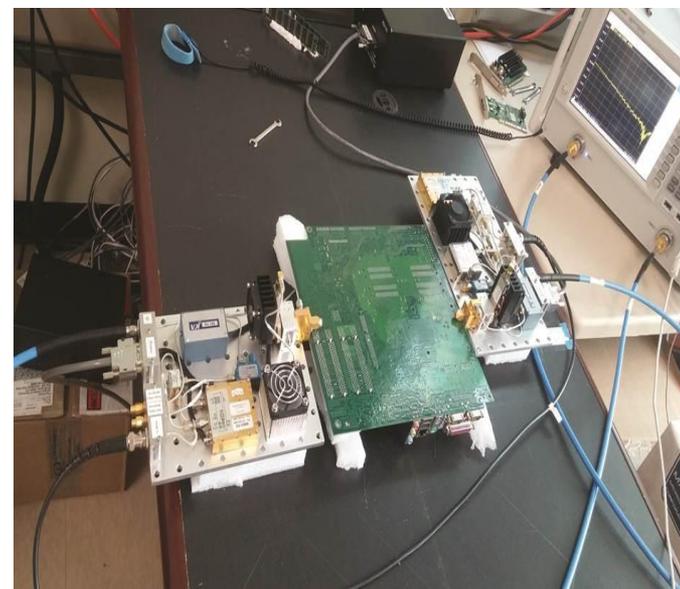
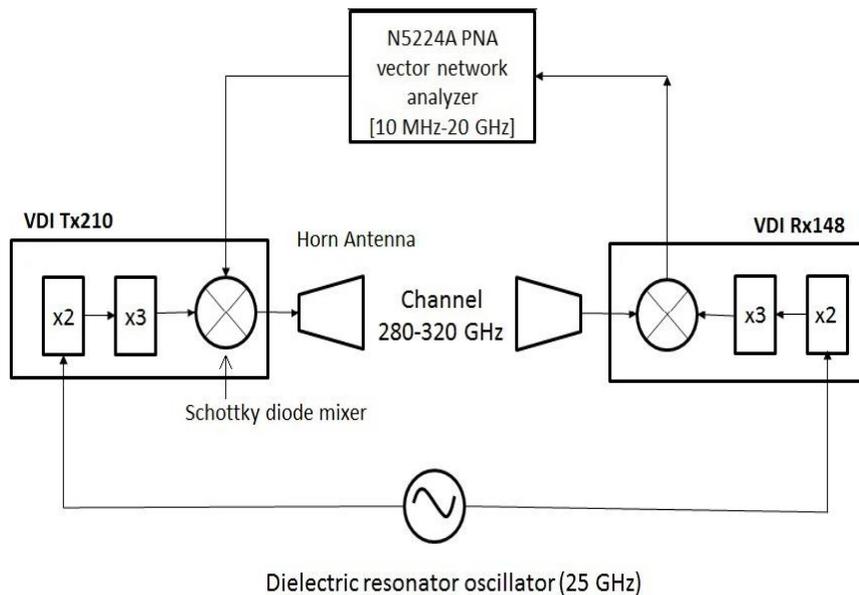
- + antenna size/integration with chips
- +adding bandwidth without adding pins or fiber connectors to the chip package.
- - limited bandwidth
- For example, a computer in a typical high performance cluster gets 56 Gbits/s
- Wireless communication at mm- Wave frequencies: WiGig uses 60 GHz frequency range to provide up to 7 Gbits/s using OFDM, 64- QAM, and sophisticated coding.
- Power hungry systems

# ❖ THz Transmitter and Receiver

## 110-170 GHz Measurement System

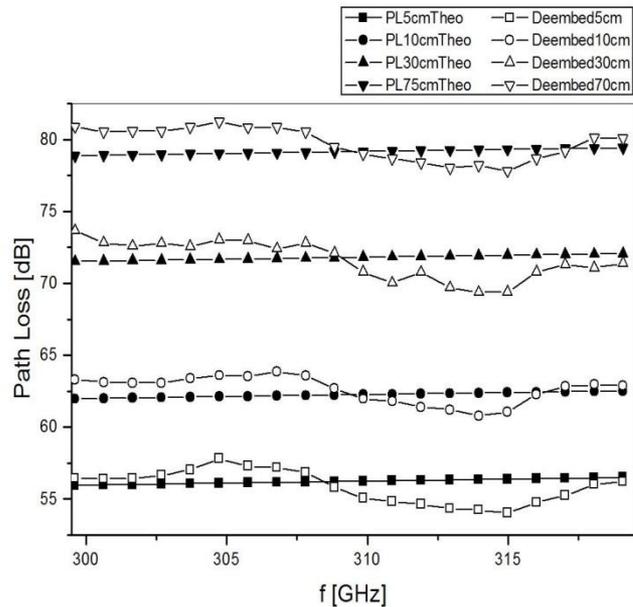


## 300-320 GHz Measurement System

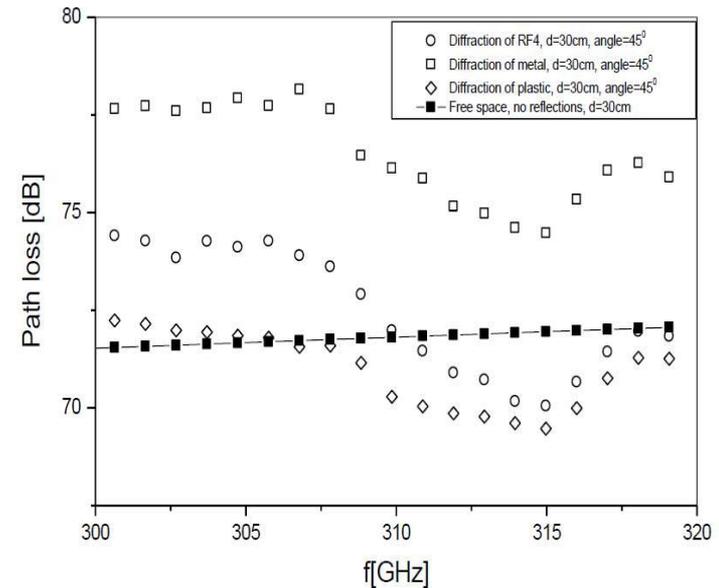


# ❖ Path Loss Measurement

## ➤ LoS



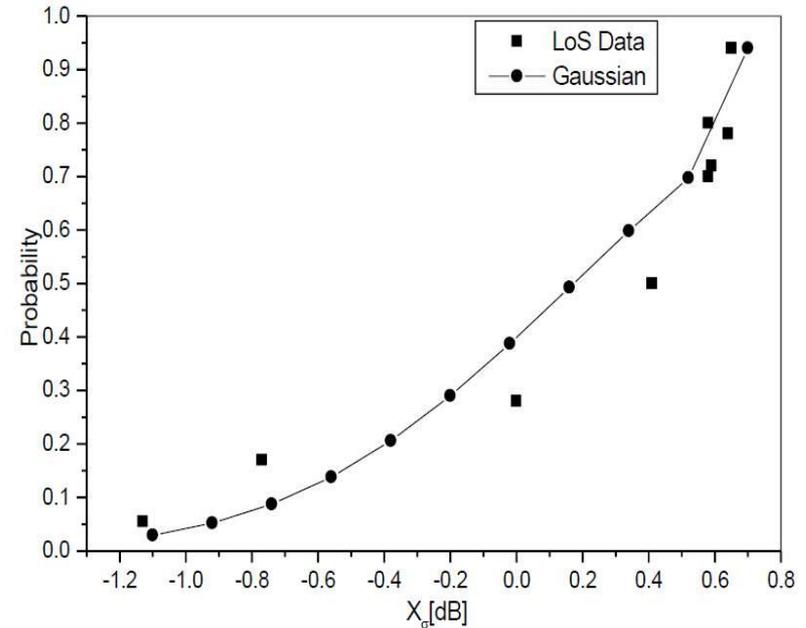
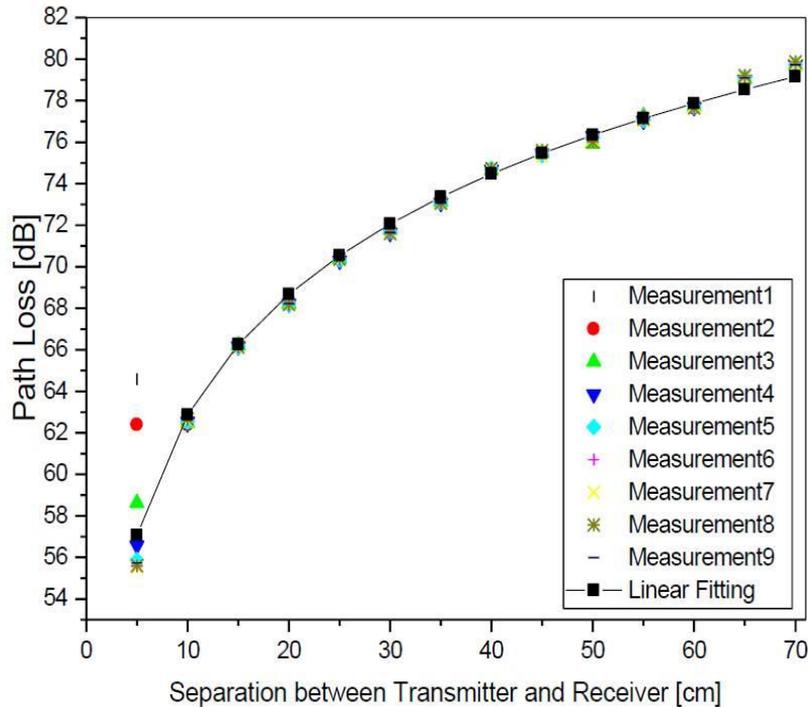
## ➤ NLoS



- Varying diffraction loss with different materials (FR4, metal, plastic)

[1] S. Kim and A. Zajić, "Statistical characterization of 300-GHz propagation on a desktop," *IEEE Transactions on Vehicular Technology*, vol. 64, no. 8, pp. 3330-3338, Aug. 2015.

# ❖ Path Loss vs. Distance



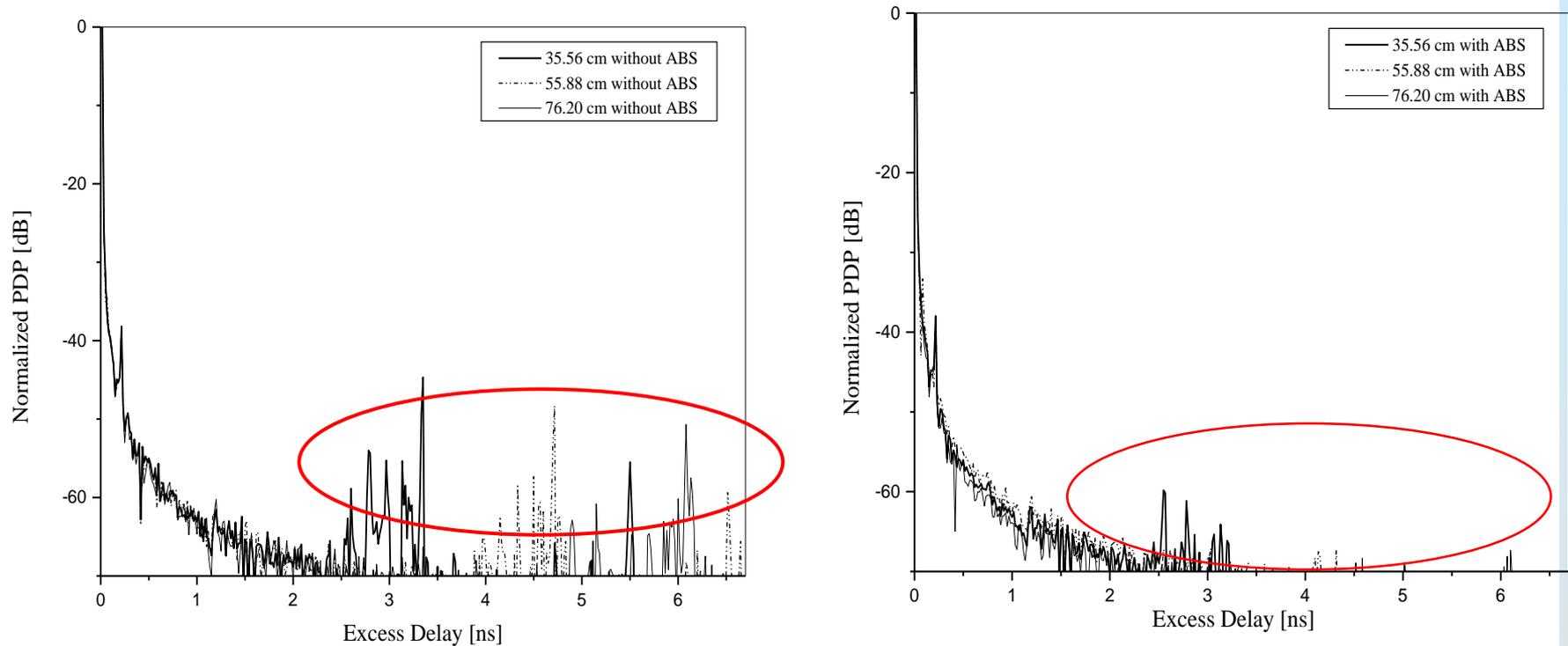
Freq. bands [GHz]	Path Loss exponent ( $\gamma$ )	$\sigma$ [dB]
300-320	1.927	0.67
300-302.5	1.916	0.67
305-307.5	1.886	0.715
310-312.5	1.93	0.737
315-317.5	1.95	0.71

- Log-normality of path loss variation introduced by misalignment:

$$PL = PL_0 + \gamma 10 \log_{10} \left( \frac{d}{d_0} \right) + X_\sigma$$

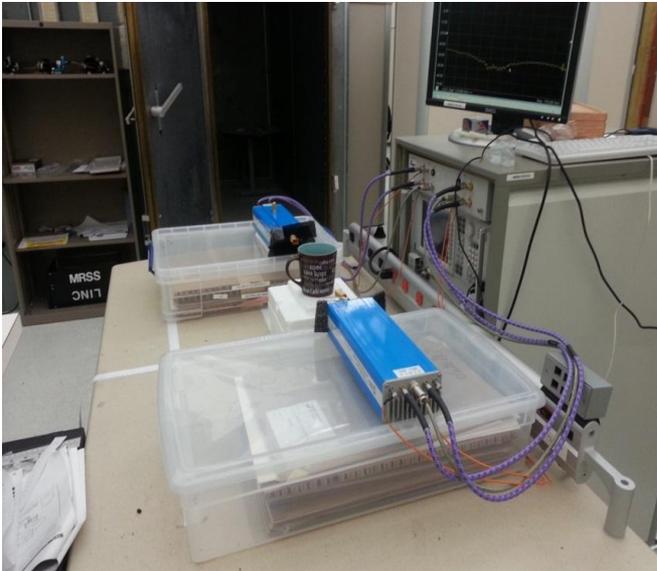
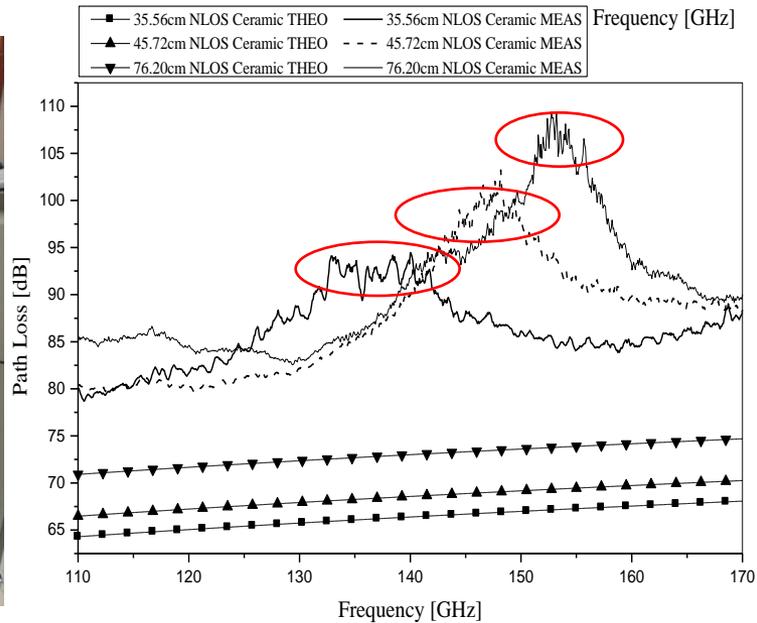
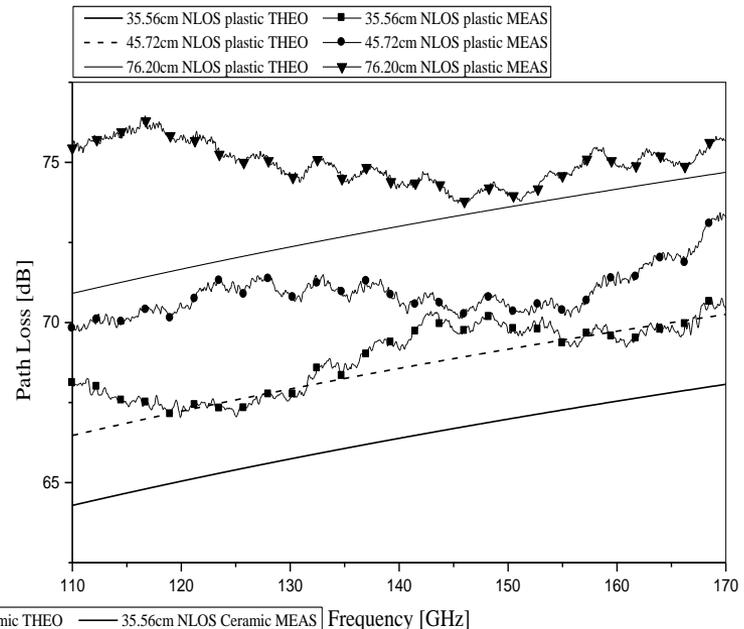
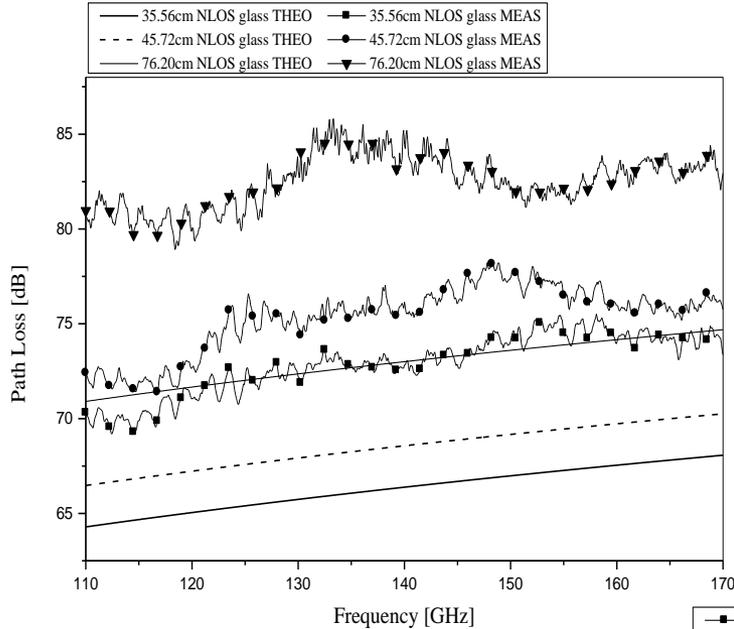
$$PL_0 = 81.97 \text{ dB} \quad (d_0 = 1 \text{ m})$$

# ❖ Multipath Characterization (LoS)

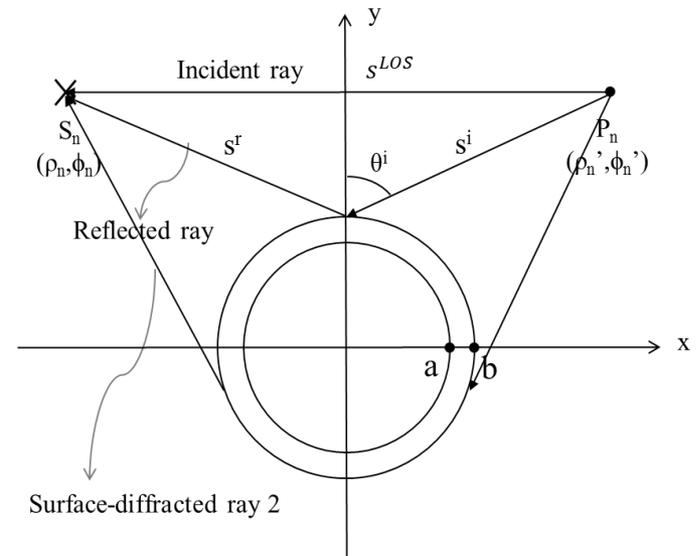
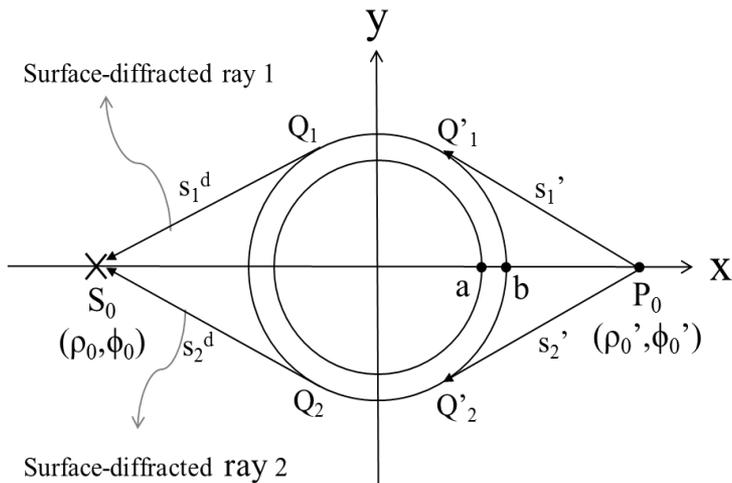
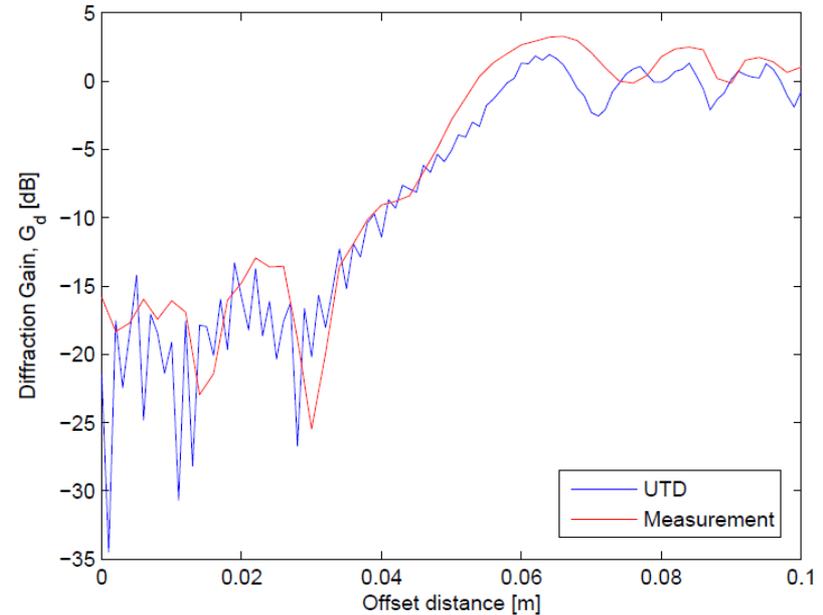
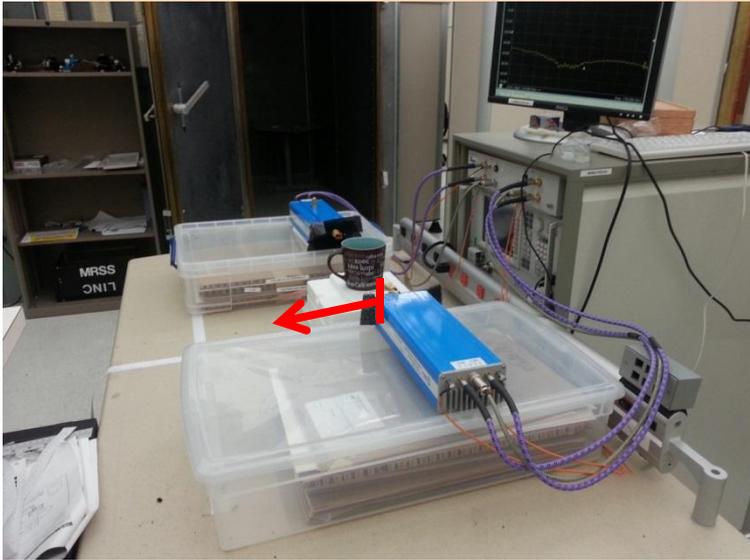


[2] S. Kim, W. T. Khan, A. Zajić, and J. Papapolymerou, "D-band channel measurements and characterization for indoor applications," *IEEE Transactions on Antennas and Propagation*, vol. 63, no. 7, pp. 3198-3207, July 2015.

# ❖ Path Loss with Cylindrical Obstructions

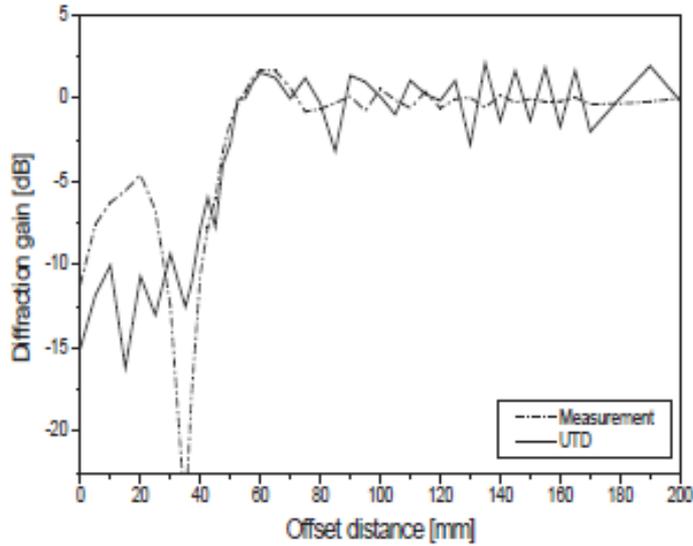


# ❖ Diffraction

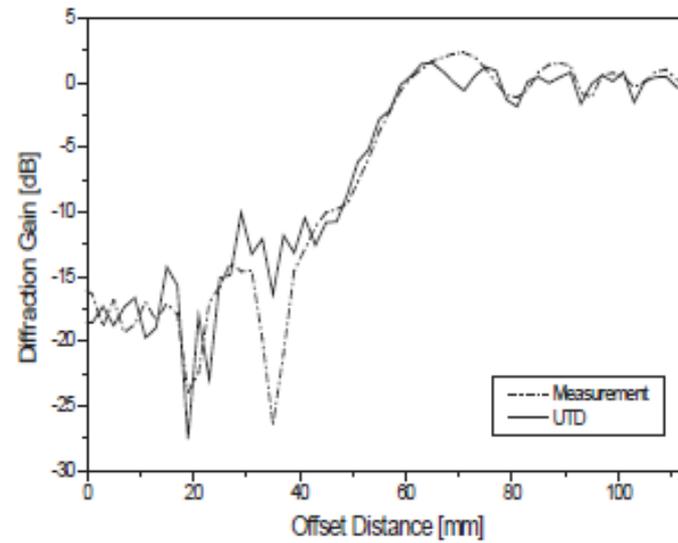


[3] S. Kim and A. Zajić, "UTD-Based Modeling of Diffraction Loss by Dielectric Circular Cylinders at D-band," *Proceedings of IEEE International Symposium on Antennas and Propagation*, pp. 1-2, June 26-July 1, 2016, Fajardo, Puerto Rico.

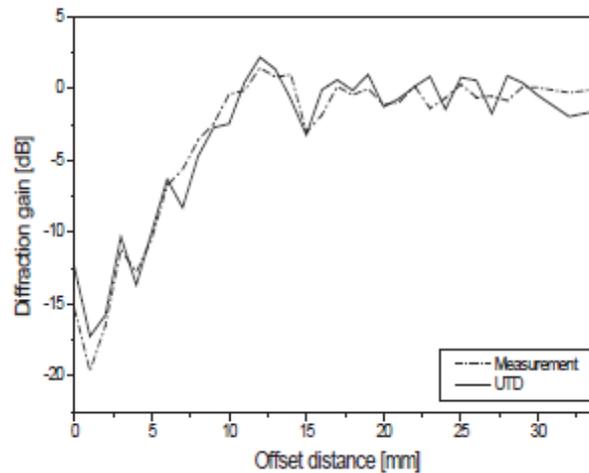
# ❖ Diffraction



33 GHz

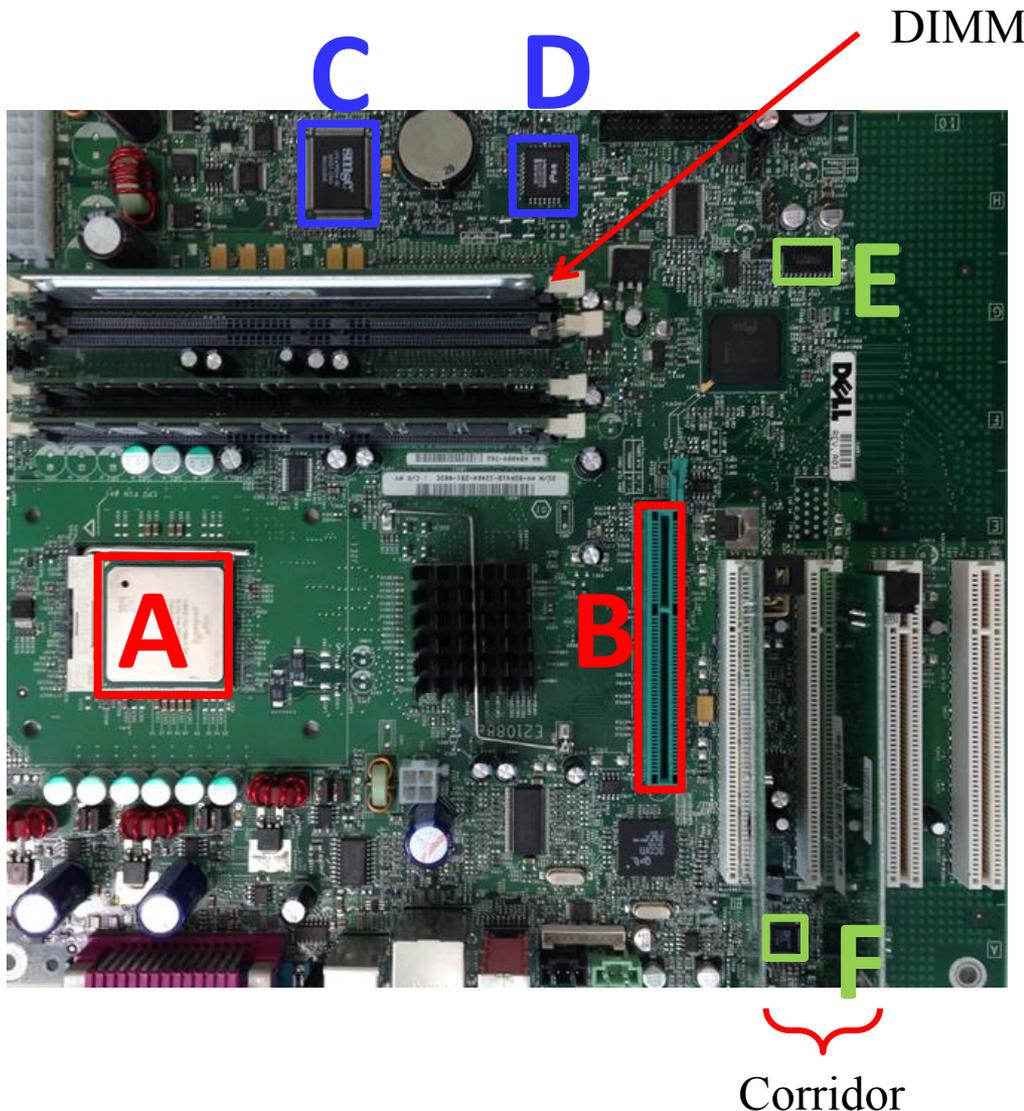


140 GHz



307 GHz

# ❖ Chip-to-Chip Measurement Scenarios



- Link A-B: CPU-AGP (Accelerated Graphics Port)  
- LoS with T-R height difference
- Link C-D: Directed NLoS with DIMM as reflecting surface
- Link E-F: OLoS through parallel-plate structure (i.e., DIMM's, cards)

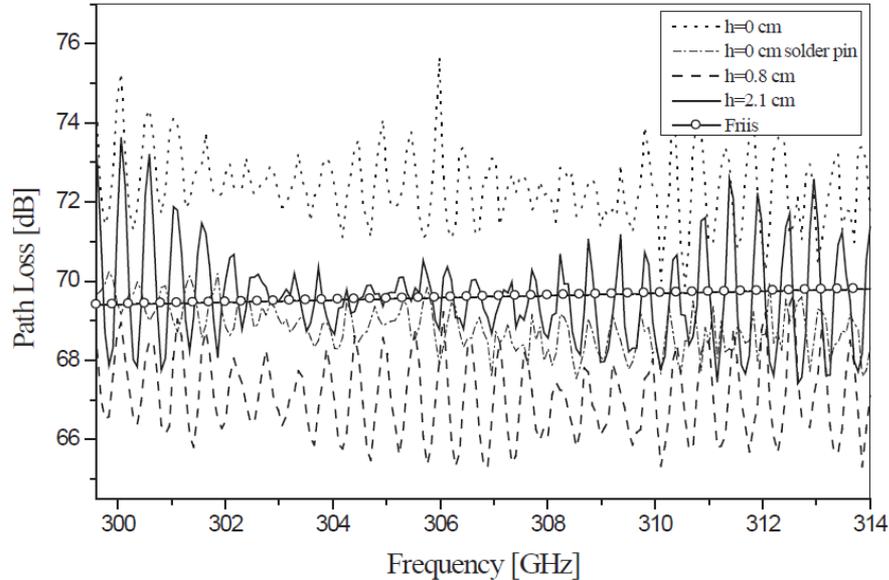
## ❖ Measurement Scenarios

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- 1) LoS propagation between the Tx and Rx over the large ground plane
- 2) Processor-Memory Link (A-B Channel)
- 3) OLoS Link through Guided Metal Parallel-plate Structures (C-D Channel)
- 4) Heatsink Channel

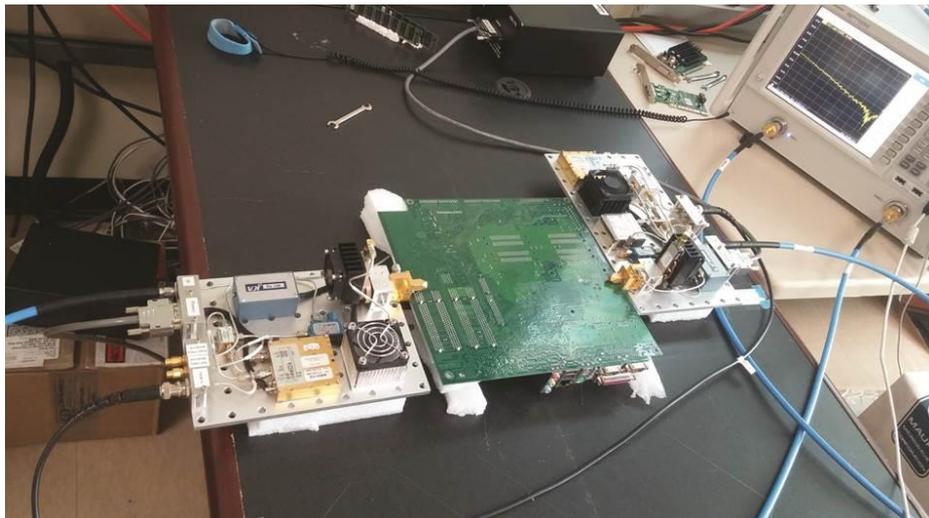
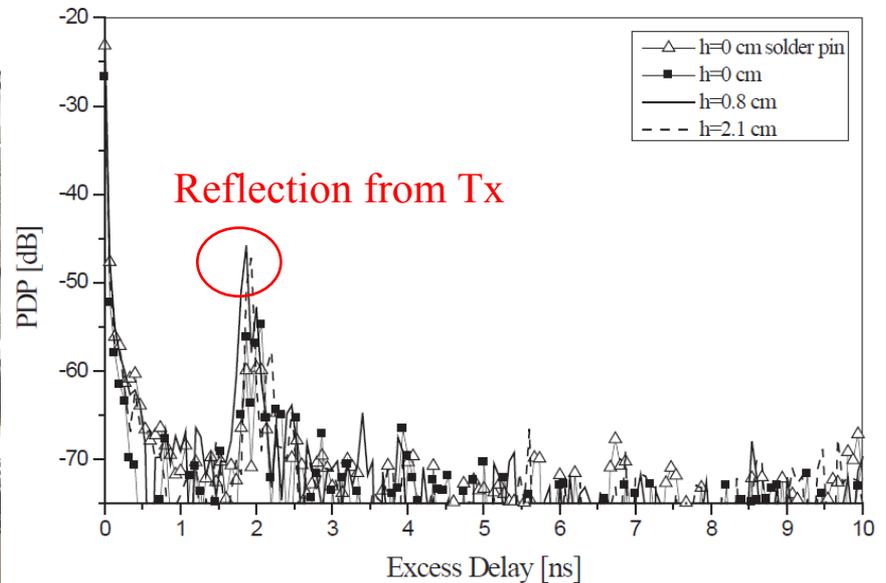
# ❖ LoS over large ground plane

## ➤ Path Loss

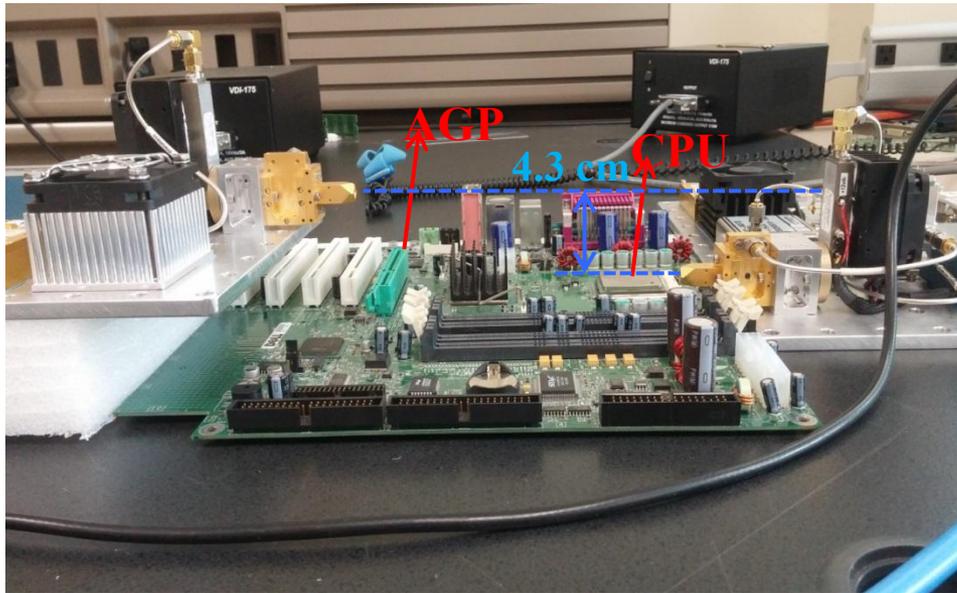


- Received power dependent on antenna height and the location of ground-reflection
- Oscillation due to reflection from Tx hardware

## ➤ PDP

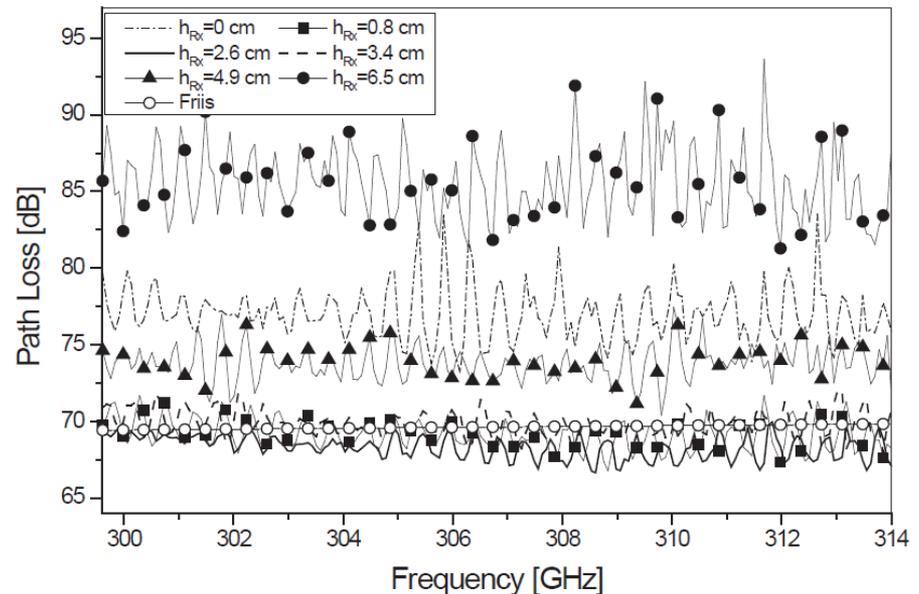


# ❖ LoS with T-R height difference

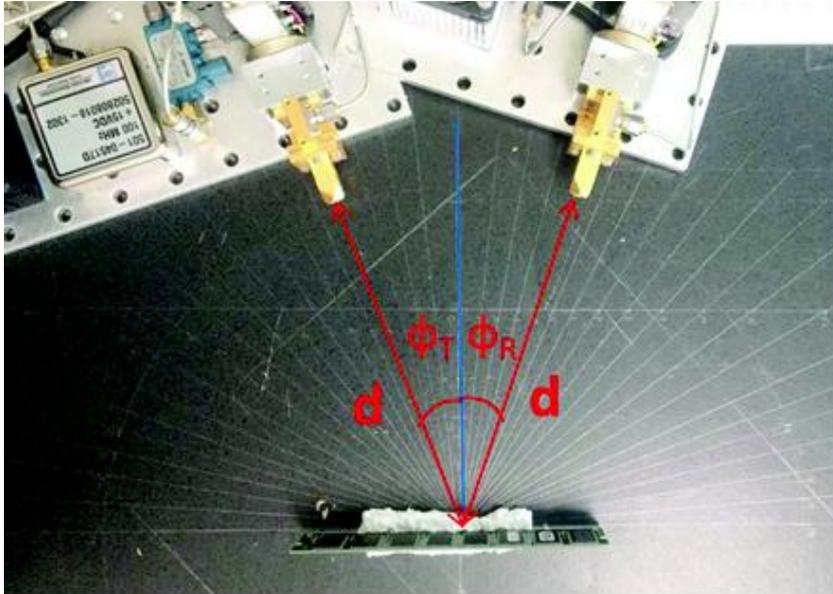


➤  $h_{Tx}=2.1$  cm,  $0$  cm  $< h_{Rx} < 6.5$  cm

- For  $\Delta h > 1.3$  cm, path loss starts deviating from theoretical value
- Height difference in the order of few centimeters ( $> 10\lambda$ ) will suffer from significant loss



# ❖ Directed NLoS



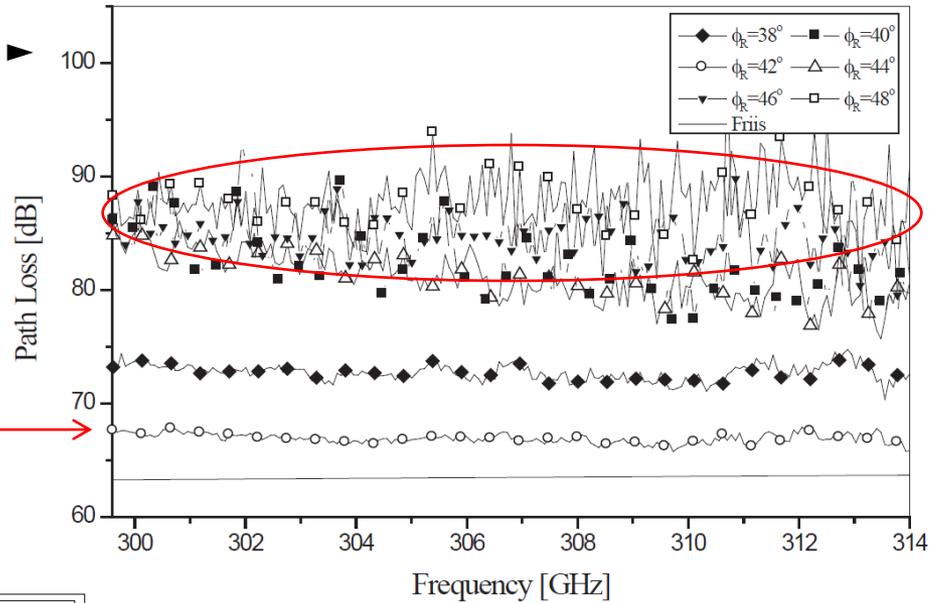
- Front and Back surfaces of a DIMM (Dual Inline Memory Module)



- Front and Back surfaces of a graphic card

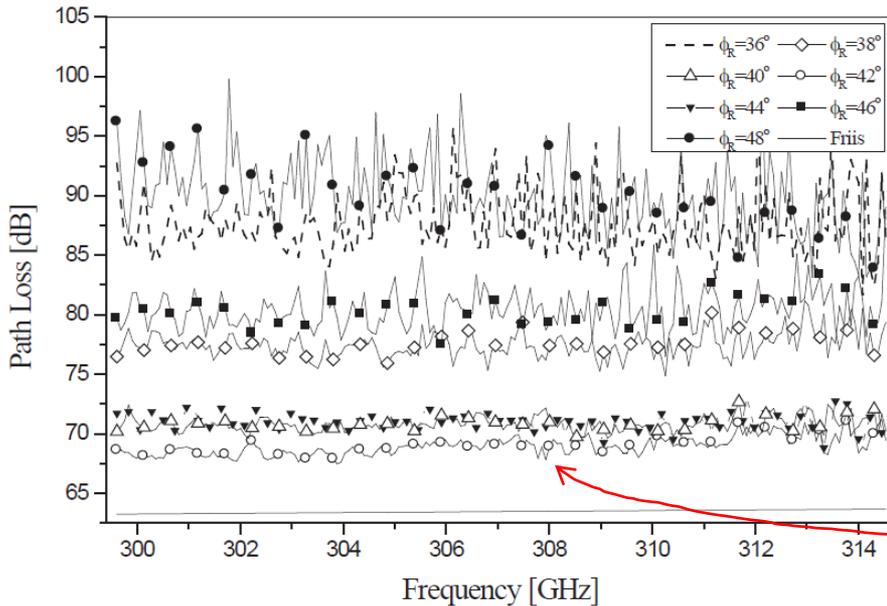
# ❖ Path Loss (Directed NLoS)

Front surface of DIMM ▶



$(\phi_T = \phi_R = 42^\circ)$  →

◀ Back surface of DIMM



↑ PL increasing linearly with the deviation from  $\phi_T$

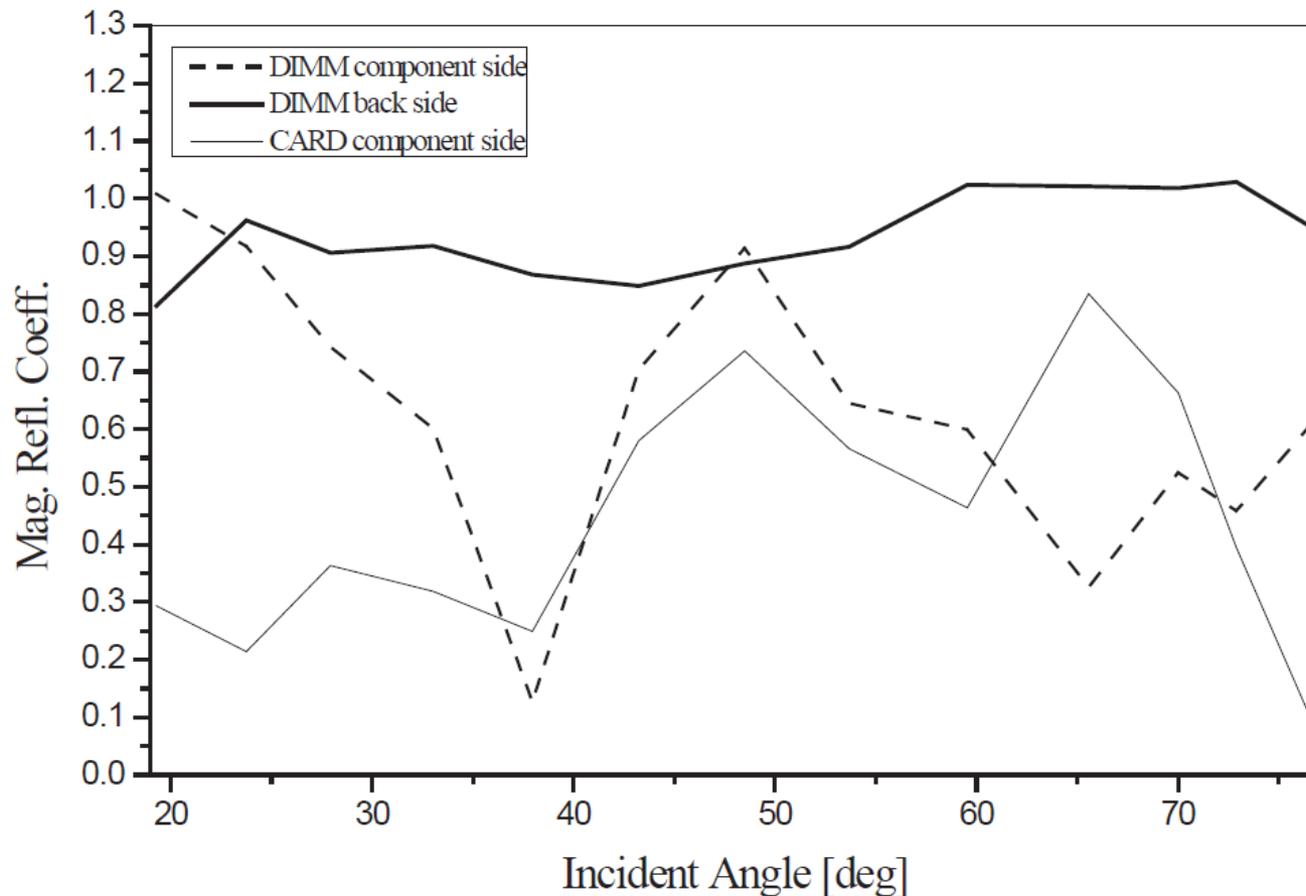
$(\phi_T = \phi_R = 42^\circ)$

## ❖ Reflection Coefficient (Directed NLoS)

$$\triangleright RL = \underbrace{P_t - P_r + G_t + G_r}_{\text{Measured PL}} - \widetilde{PL}$$

$$|\Gamma| = 10^{-RL/20}$$

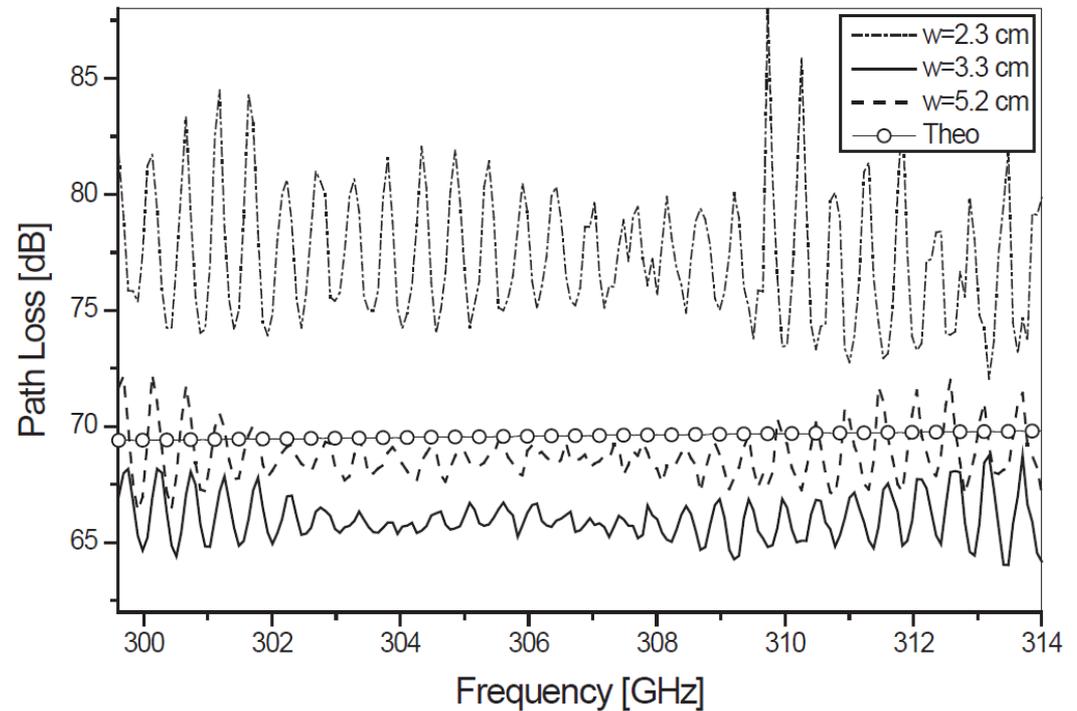
*Measured PL*



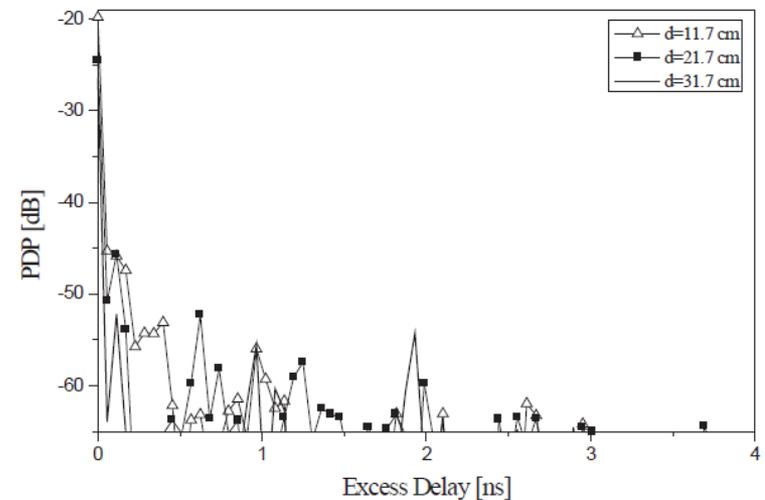
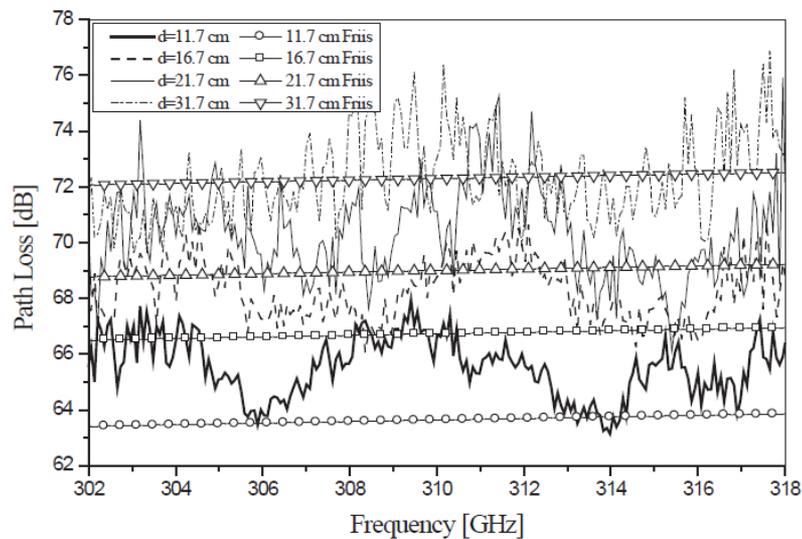
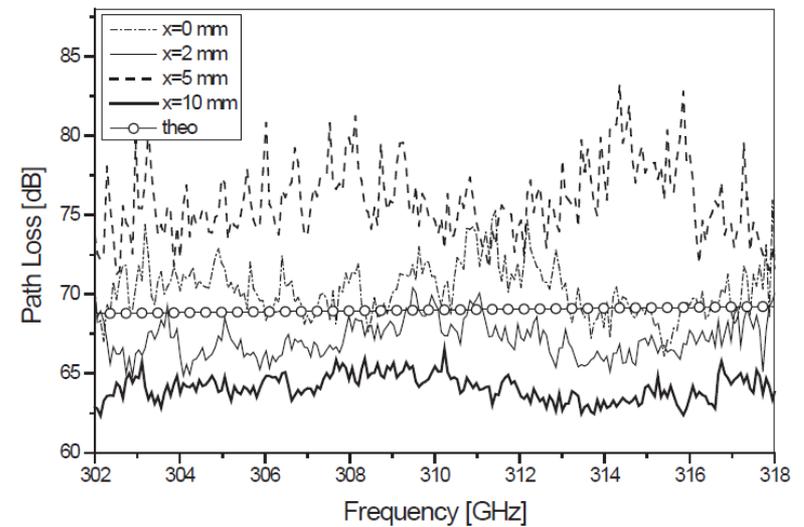
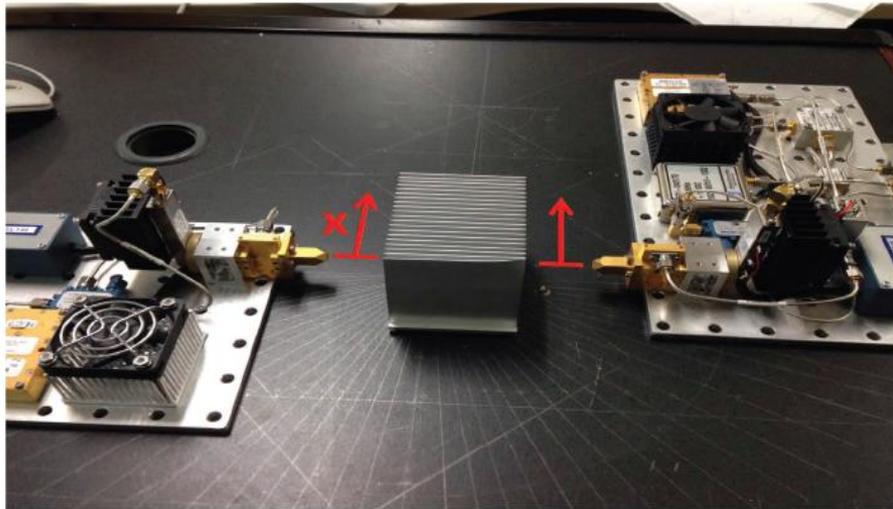
# ❖ OLoS Link through Parallel-Plate Structure



➤  $2.3 \text{ cm} < w < 5.2 \text{ cm}$

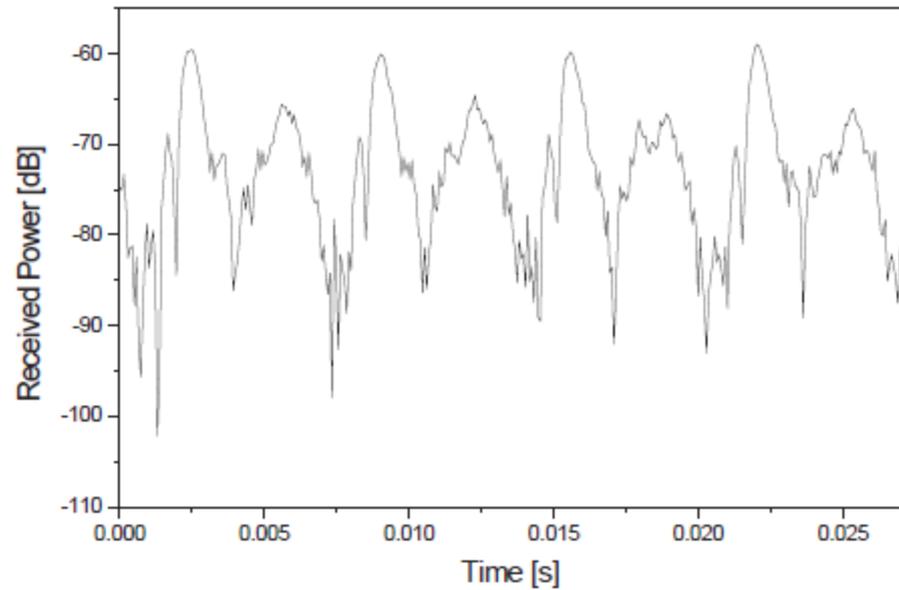
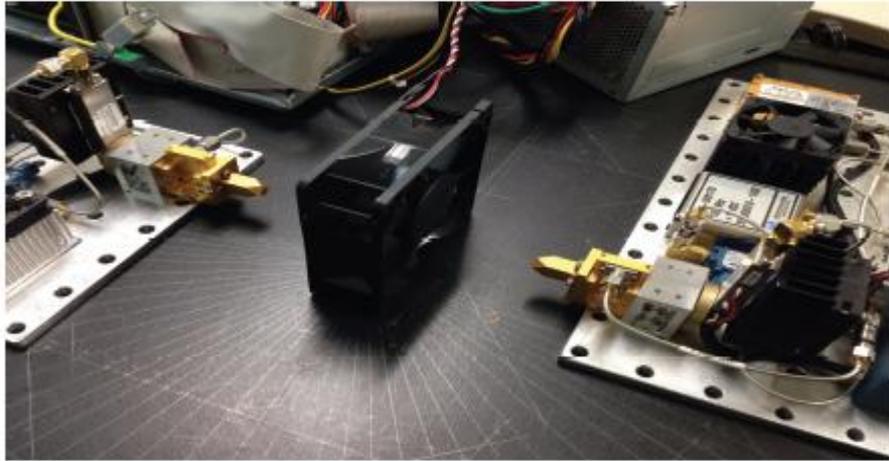


# ❖ NLOS- Through Heatsink

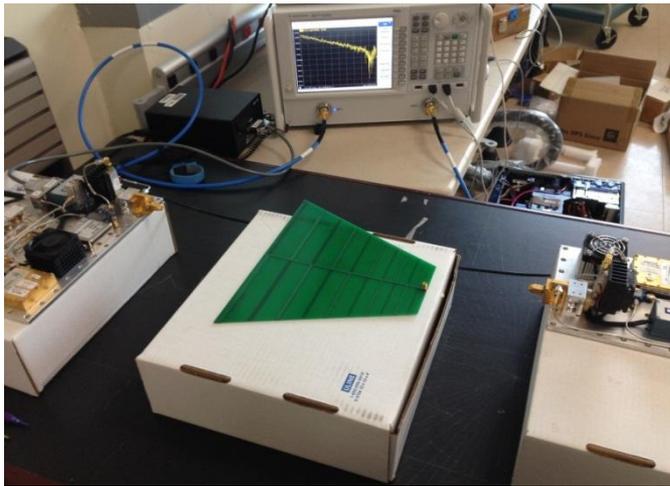


# ❖ NLOS- Through Rotating Fan

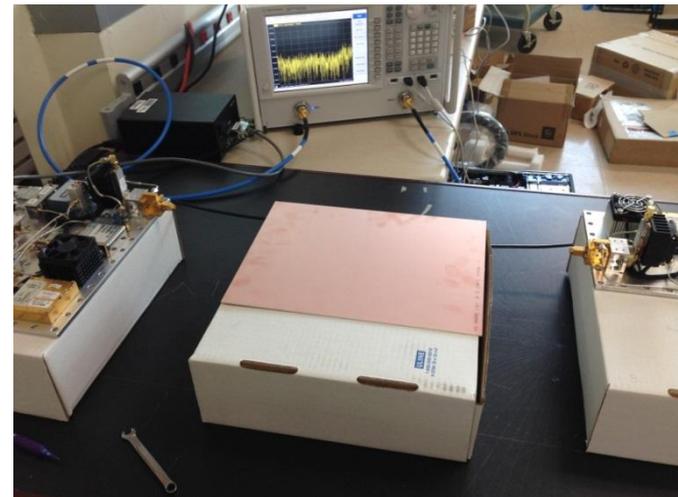
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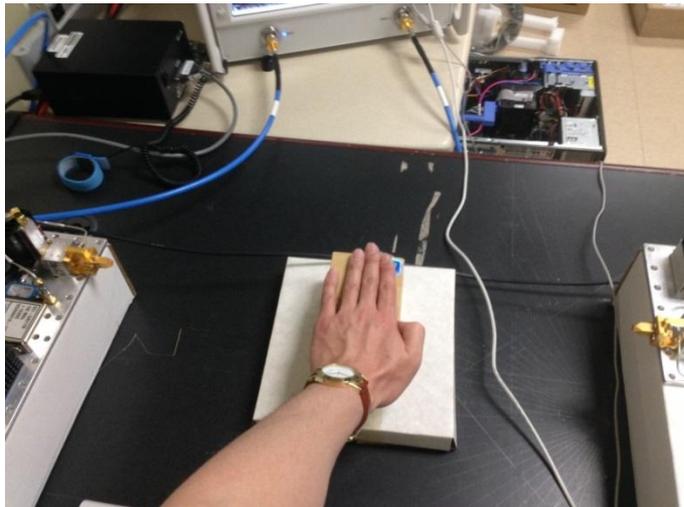
# ❖ Impact of Human Hand on THz Propagation



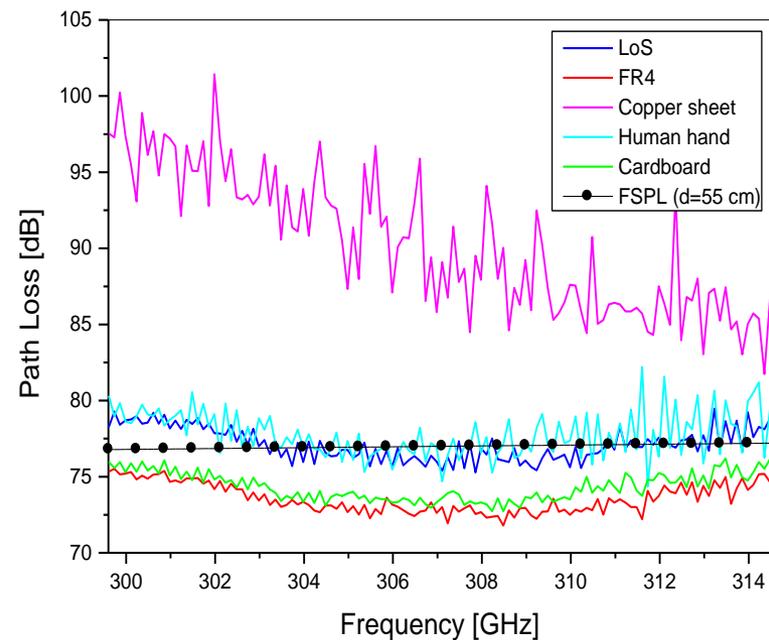
Ground reflection (FR4)



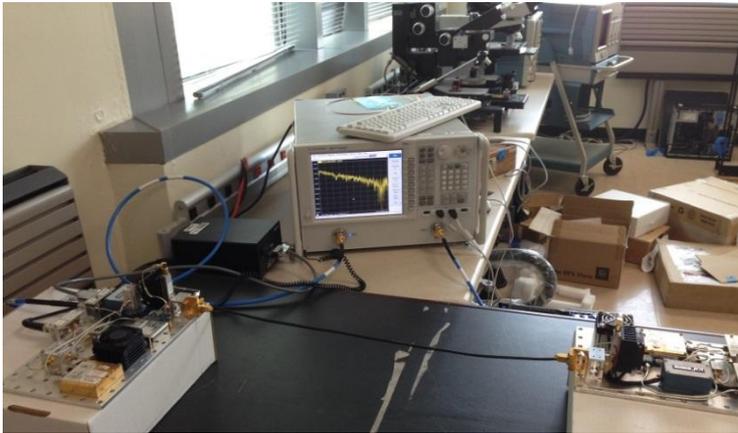
Ground reflection (Copper sheet)



Ground reflection (Human hand)

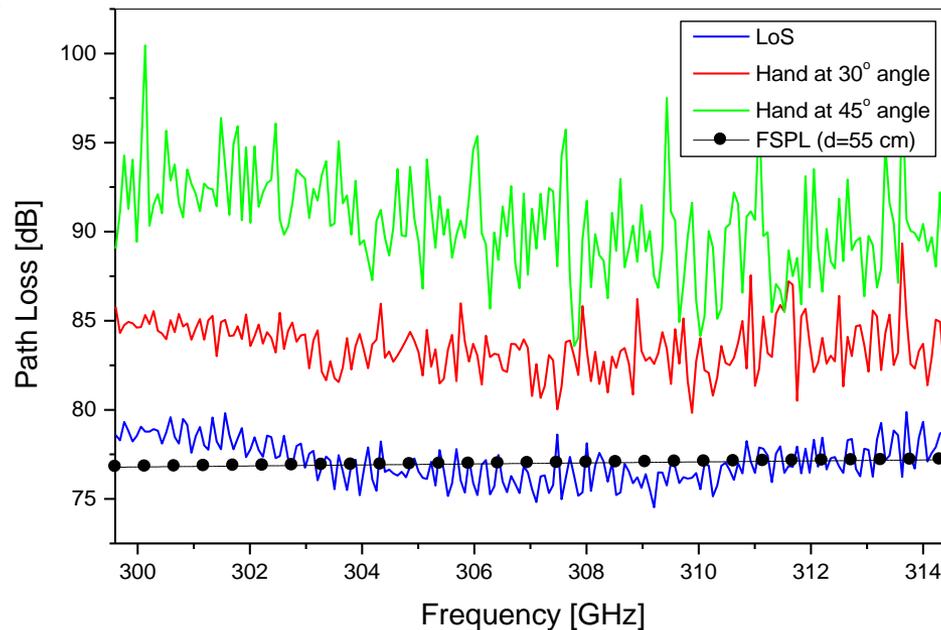


# ❖ Impact of Human Hand on THz Propagation

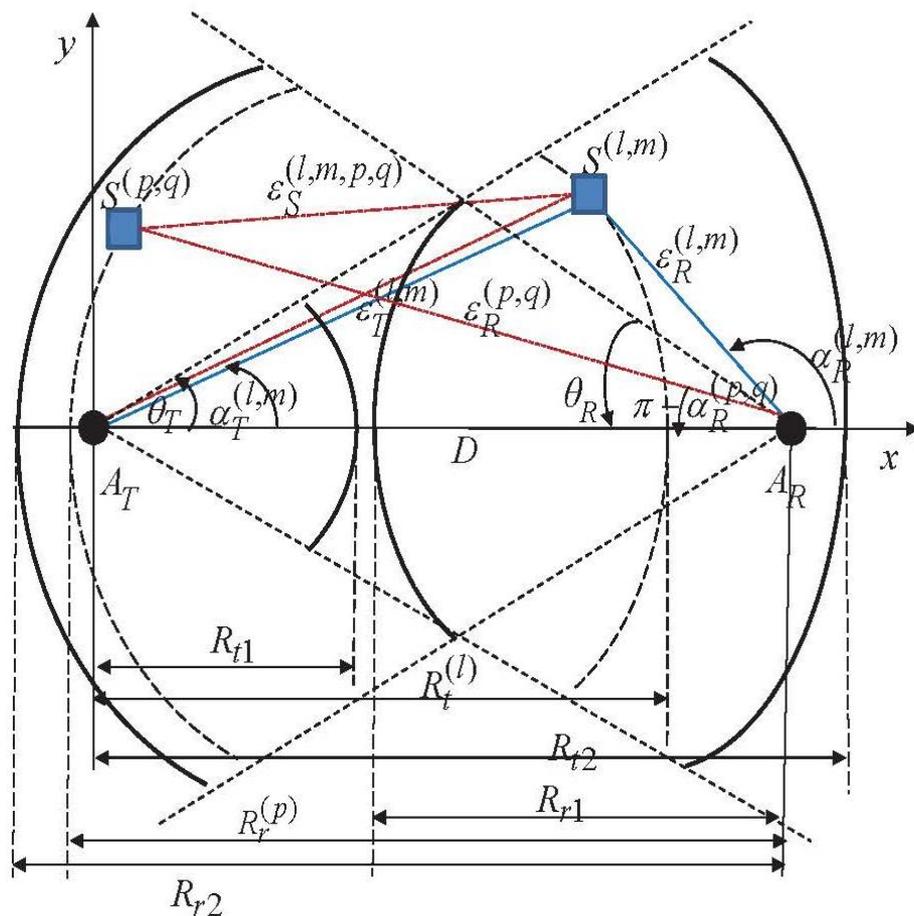


Perturbation by hand

LoS



## ❖ 2-D Geometrical Propagation Model

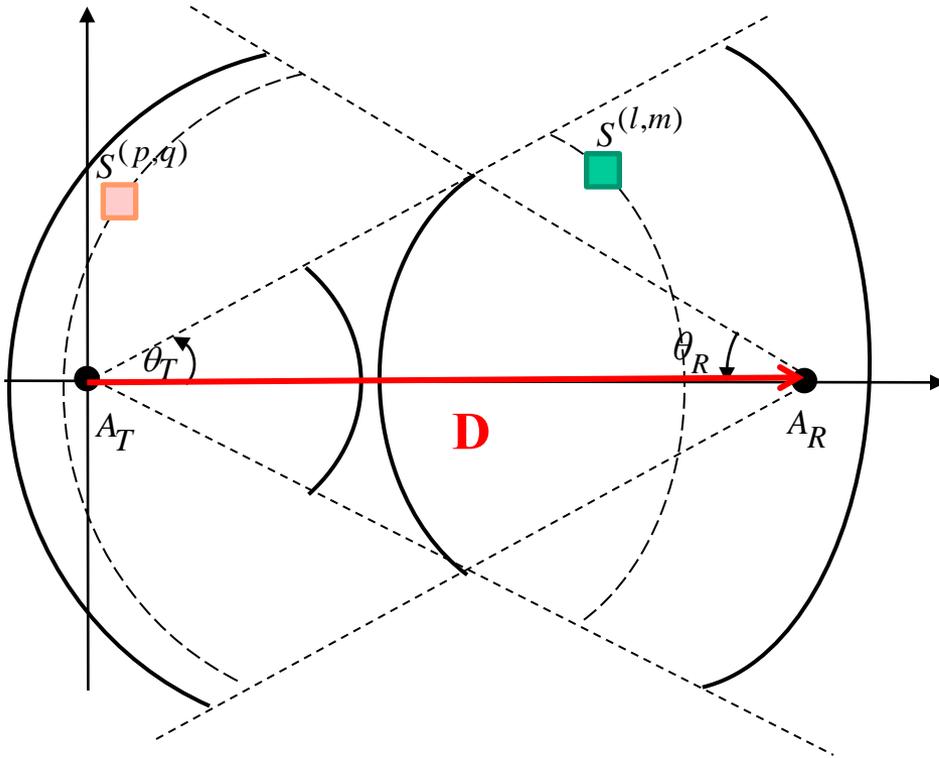


- Three components:
  - LoS
  - Single-Reflected
  - Double-Reflected

- Quasi-static channel  
→ Time-invariant delay-spread function,  $h(\tau)$

[5] S. Kim and A. Zajić, "Statistical modeling and simulation of short-range device-to-device communication channels at sub-THz frequencies," *IEEE Transactions on Wireless Communications*, vol. 15, no. 9, pp. 6423 – 6433, Sept. 2016

## ❖ LoS Ray

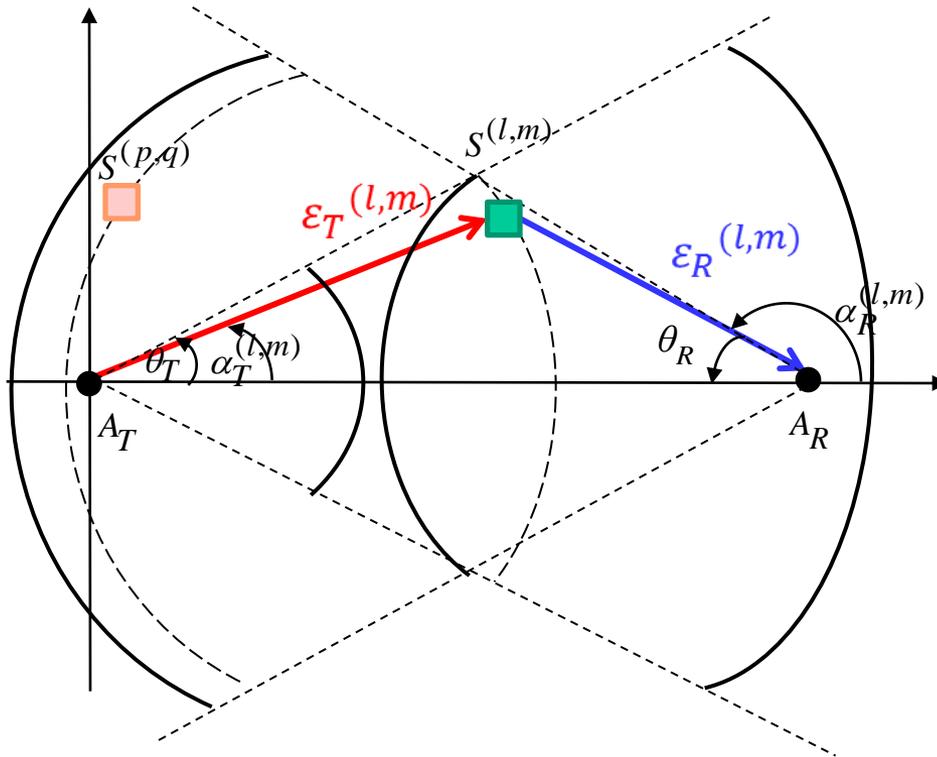


$$A_{LoS} = \frac{\lambda \sqrt{G_T G_R}}{4\pi D \gamma/2}$$

$$\tau_{LoS} = \frac{D}{c_0}$$

$$h^{LoS}(\tau) = \sqrt{\frac{K}{K+1}} A_{LoS} e^{j\phi_{LoS}} \delta(\tau - \tau_{LoS})$$

## ❖ Single-Reflected Ray



$$A_{l,m} = \frac{\lambda \sqrt{G_T G_R}}{4\pi (\epsilon_T^{(l,m)} + \epsilon_R^{(l,m)})^{3/2}}$$

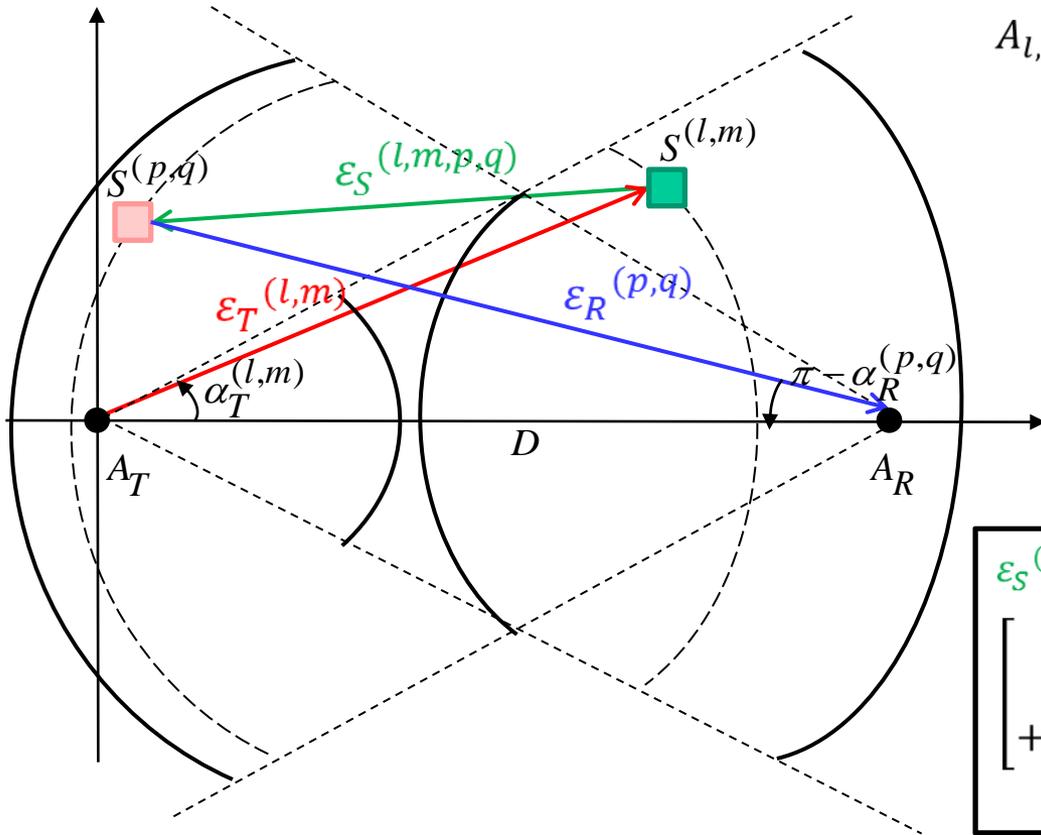
$$\tau_{l,m} = \frac{\epsilon_T^{(l,m)} + \epsilon_R^{(l,m)}}{c_0}$$

$$\epsilon_T^{(l,m)} = R_t^{(l)}$$

$$\epsilon_R^{(l,m)} = \frac{R_t^{(l)} \cos \alpha_T^{(l,m)} - D}{\cos \left[ \tan^{-1} \left( -\frac{R_t^{(l)} \sin \alpha_T^{(l,m)}}{D - R_t^{(l)} \cos \alpha_T^{(l,m)}} \right) + \pi \right]}$$

$$h^{SR}(\tau) = \sqrt{\frac{\eta_{SR}}{K+1}} \lim_{M \rightarrow \infty} \frac{1}{\sqrt{M}} \sum_{l=1}^L \sum_{m=1}^{M^{(l)}} A_{l,m} e^{j\phi_{l,m}} \delta(\tau - \tau_{l,m})$$

# ❖ Double-Reflected Ray



$$A_{l,m,p,q} = \frac{\lambda \sqrt{G_T G_R}}{4\pi (\epsilon_T^{(l,m)} + \epsilon_R^{(l,m)} + \epsilon_S^{(l,m,p,q)})^{3/2}}$$

$$\tau_{l,m,p,q} = \frac{\epsilon_T^{(l,m)} + \epsilon_R^{(l,m)} + \epsilon_S^{(l,m,p,q)}}{c_0}$$

$$\epsilon_T^{(l,m)} = R_t^{(l)}$$

$$\epsilon_R^{(p,q)} = R_r^{(p)}$$

$$\epsilon_S^{(l,m,p,q)} = \left[ |R_t^{(l)} \sin \alpha_T^{(l,m)} - R_r^{(p)} \sin \alpha_R^{(p,q)}|^2 + |R_t^{(l)} \cos \alpha_T^{(l,m)} - R_r^{(p)} \cos \alpha_R^{(p,q)} - D|^2 \right]^{1/2}$$

$$h^{DR}(\tau) = \sqrt{\frac{\eta_{DR}}{K+1}} \lim_{M,Q \rightarrow \infty} \frac{1}{\sqrt{MQ}} \sum_{l,m=1}^{L,M^{(l)}} \sum_{p,q=1}^{P,Q^{(p)}} A_{l,m,p,q} e^{j\phi_{l,m,p,q}} \delta(\tau - \tau_{l,m,p,q})$$

# ❖ Distribution of Scatterers

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- AoD's ( $\alpha_T^{(l,m)}$ ), AoA's, ( $\alpha_R^{(p,q)}$ ), and sector radii ( $R_t^{(l)}$ ,  $R_r^{(p)}$ ) are independent and uniformly distributed random variables.
  - Such a distribution implies that the scatterers will have a uniform density between the concentric-sectors, if the scattering is isotropic
- The joint PDF:

$$f(R, \alpha) = f(R) \cdot f(\alpha) = \frac{2R}{(\alpha_2 - \alpha_1)(R_2^2 - R_1^2)}$$

- Phases ( $\phi_{LoS}$ ,  $\phi_{l,m}$ ,  $\phi_{l,m,p,q}$ ) are uniformly distributed random variables on the interval  $[-\pi, \pi)$ , and independent from any other R.V.'s

## ❖ Correlation Function

---

- Transfer function is the FFT pair of the delay-spread function

$$T(f) = \mathcal{F}\{h(\tau)\} = T^{LoS}(f) + T^{SR}(f) + T^{DR}(f)$$

- Frequency Correlation Function (FCF) is a measure of the channel's frequency selectivity; an indicator of a required frequency difference ( $\Delta f$ ) between sample points for the values to be effectively uncorrelated

$$R(\Delta f) = \frac{E[T^*(f) T(f + \Delta f)]}{\sqrt{VAR[T^*(f)] VAR[T(f + \Delta f)]}}$$

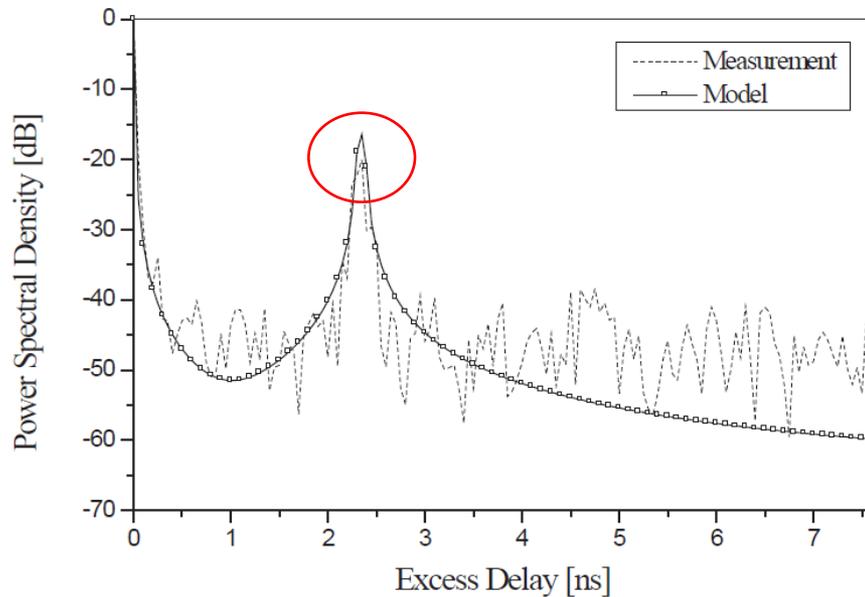
- Since  $T^{LoS}(f)$ ,  $T^{SR}(f)$ ,  $T^{DR}(f)$  are independent zero-mean complex Gaussian random processes,

$$R(\Delta f) = R^{LoS}(\Delta f) + R^{SR}(\Delta f) + R^{DR}(\Delta f)$$

# ❖ Reference Model Validation

## ➤ LoS desktop scenario (300~320 GHz) :

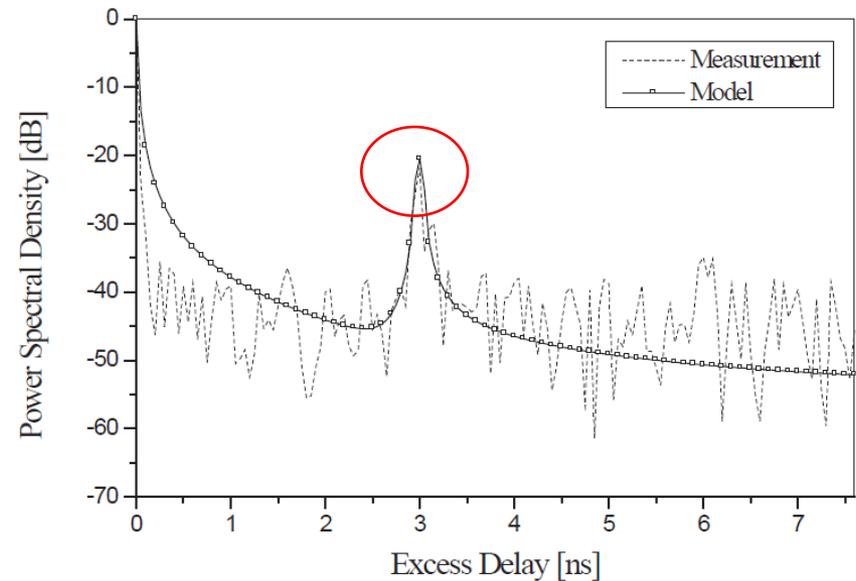
$$2\theta_T = 2\theta_R = 10^\circ, \eta_{SR} = 0, \eta_{DR} = 1, K = 0.4$$



$$D = 30\text{cm}$$

$$R_{t1} = R_{r1} = 32\text{cm}$$

$$R_{t2} = R_{r2} = 33\text{cm}$$



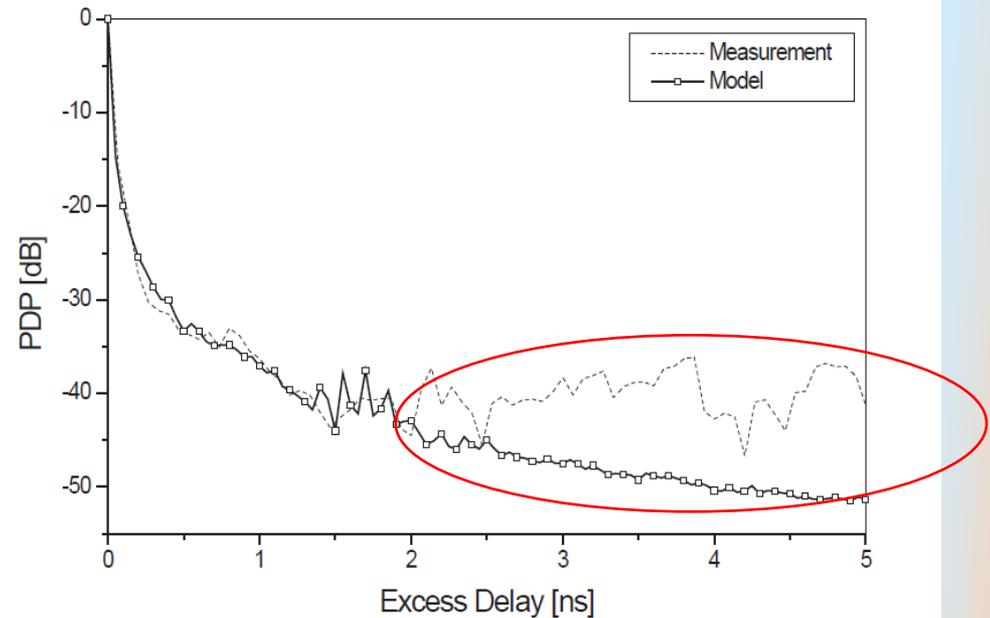
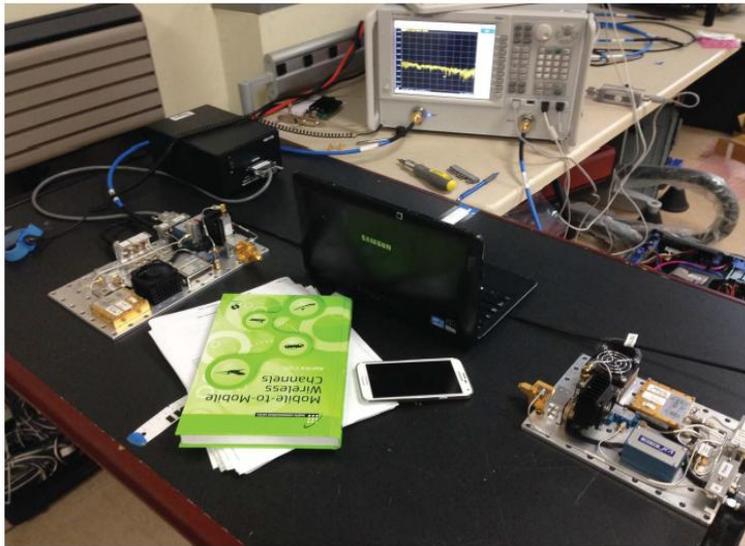
$$D = 40\text{cm}$$

$$R_{t1} = R_{r1} = 42\text{cm}$$

$$R_{t2} = R_{r2} = 43\text{cm}$$

# ❖ Reference Model Validation

- Realistic desktop scenario with clutters (300~320 GHz) :



$$\eta_{SR} = 0.3, \eta_{DR} = 0.7,$$

$$K = 0.5$$

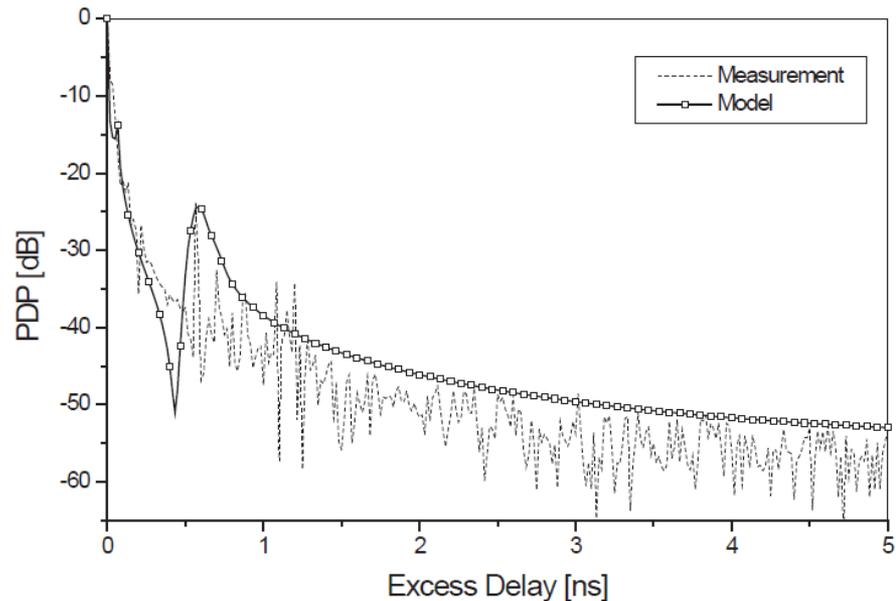
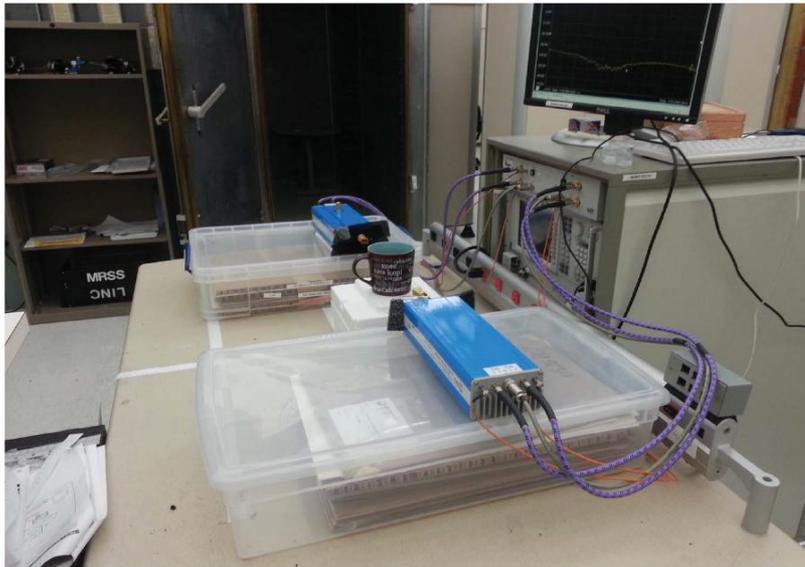
$$D = 55\text{cm}$$

$$R_{t1} = R_{r1} = 10\text{cm}$$

$$R_{t2} = R_{r2} = 50\text{cm}$$

# ❖ Reference Model Validation

- NLoS desktop scenario with cylindrical obstruction (110~170 GHz) :



$$\eta_{SR} = \eta_{DR} = 0.5,$$

$$K = 0.15$$

$$D = 35.56 \text{ cm}$$

$$R_{t1} = R_{r1} = 21.5 \text{ cm}$$

$$R_{t2} = R_{r2} = 22.5 \text{ cm}$$

# ❖ Research Challenges

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- Cost and energy efficient transceivers
- Antenna design for efficient communication over small distances
- Channel modeling at THz frequencies
- Low-complexity modulation and coding schemes
- Channel equalization over wide frequency bandwidth
- Medium access protocols suitable for ultra-dense networks

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THANK YOU

Questions?