## Project: IEEE P802.15 Working Group for Wireless Personal Area Networks (WPANs)

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Abstract: [Propose SFD values and FEC schemes for 802.15 .4 g standard]
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## Summary

- This document presents two 16-bit SFDs, and convolution code ( $R=1 / 2, K=4$ ).
- SFD: 3 plans are available for FEC and Non-FEC identification

| Plan | SFD Value for FEC | SFD value for Non-FEC |
| :---: | :---: | :---: |
| A | $0 \times F 68 D$ | $0 \times 7 B C 9$ |
| B | $0 \times 6 F 4 E$ | $0 \times 904 E$ |
| C | $0 \times 21 F 6$ | $0 \times C 9 C 2$ |

- FEC:

| Mode | $\mathbf{R}$ | $\mathbf{m}$ | $\mathbf{n}$ | $\mathbf{k}$ | $\mathbf{L}$ | $\mathbf{g 0}$ | $\mathbf{g 1}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Systematic | $1 / 2$ | 3 | 2 | 1 | 4 | $\left\{\begin{array}{lllll}1 & 1 & 1 & 1\end{array}\right\}$ | $\left\{\begin{array}{llll}1 & 1 & 0 & 1\end{array}\right\}$ |
| Non Systematic | $1 / 2$ | 3 | 2 | 1 | 4 | $\left\{\begin{array}{llll}1 & 1 & 1 & 1\end{array}\right\}$ | $\left\{\begin{array}{llll}1 & 1 & 0 & 1\end{array}\right\}$ |

## Proposal on SFD

## Proposal on SFD

| Plan | SFD Value for FEC mode | SFD value for Non-FEC mode |
| :---: | :---: | :---: |
| A | $0 \times F 68 D$ | $0 \times 7 B C 9$ |
| B | $0 \times 6 F 4 E$ | $0 \times 904 E$ |
| C | $0 \times 21 F 6$ | $0 \times C 9 C 2$ |

- Reasons to select the SFD values
- Plan A: Good auto- and cross- correlation values with moderate peak in 15.4d SHR
- Plan B: Good FA (false alarm) and MD (Miss detection) probabilities with moderate peak in 15.4d SHR
- Plan C: Better FA and MD probabilities than option B with higher peak in 15.4d SHR
- Note: Shall attain lower FA and MD probabilities when FEC is used for payload
- Topic to be clarified: Robustness of the selected SFD when FEC is used
- Should attain robustness so that the performance is negligible against payload performance


## Correlation performance of Plan A ( $a=0 x F 68 D, b=0 x 7 B C 9$ )

- Left figs: Correlation values between a and 15.4 g preamble ( 24 bit ) $+\mathrm{a},-\mathrm{a}, \mathrm{b}$, or -b
- Right figs: Correlation values between $b$ and 15.4 g preamble ( 24 bit ) $+\mathrm{a},-\mathrm{a}, \mathrm{b}$, or-b
- Note: maximal correlation values between 15.4d-preamble+15.4d-SFD and a, -a, b, or -b are 6


Auto correlation image a, SidePeak $=4$, SideRms=2



Cross correlation image b with a, SidePeak=4,SideRms=2.0736


Auto correlation b, SidePeak=2,SideRms=1.1094




Cross correlation image a with b, SidePeak=6,SideRms=2.0976


## MD and FA probabilities (Pm and Pf) Plan A (a=0xF68D, b=0x7BC9) at Eb/N0=3dB

From left figure, the optimum threshold is equal to 10 . Then, we can achieve error rate of $1.3 \times 10^{-3}$. If PER performance of payload is higher than $1.3 \times 10^{-3}$ at $\mathrm{Eb} / \mathrm{No}=3 \mathrm{~dB}$ the selected SFD is robust enough not to affect payload performance.

- Correlator values: a

- Correlator values: b

$\operatorname{Pf}(x y)$ denotes that Input signal in correlators is Preamble $+x$ and correlator values are $y$


## Correlation performance of Plan B ( $\mathrm{a}=0 \times 6 \mathrm{~F} 4 \mathrm{E}, \mathrm{b}=0 \times 904 \mathrm{E}$ )

- Left figs: Correlation values between a and 15.4 g preamble ( 24 bit ) $+\mathrm{a},-\mathrm{a}, \mathrm{b}$, or -b
- Right figs: Correlation values between $b$ and 15.4 g preamble ( 24 bit ) $+\mathrm{a},-\mathrm{a}, \mathrm{b}$, or-b
- Note: maximal correlation values between 15.4d-preamble+15.4d-SFD and a, -a, b, or -b are 4



## MD and FA probabilities (Pm and Pf) -Plan B (a=0x6F4E, $b=0 \times 904 E)$ at $\mathrm{Eb} / \mathrm{NO}=3 \mathrm{~dB}$ -

From the left figure, the optimum detection threshold is equal to 10 . Then, we can achieve error rate of $1.1 \times 10^{-3}$. If PER performance of payload is higher than $1.1 \times 10^{-3}$ at $\mathrm{Eb} / \mathrm{No}=3 \mathrm{~dB}$, the selected SFD is robust enough not to affect payload performance.

- Correlator values: a

- Correlator values: b

$\operatorname{Pf}(x y)$ denotes that Input signal in correlators is Preamble $+x$ and correlator values are $y$


## Correlation performance of Plan C(a=0x21F6, b=0xC9C2)

- Left figs: Correlation values between a and 15.4 g preamble ( 24 bit ) $+\mathrm{a},-\mathrm{a}, \mathrm{b}$, or -b
- Right figs: Correlation values between $b$ and 15.4 g preamble ( 24 bit ) $+\mathrm{a},-\mathrm{a}, \mathrm{b}$, or-b
- Note: maximal correlation values between 15.4d-preamble+15.4d-SFD and a, -a, b, or -b are 6


Cross correlation image $b$ with $a$, SidePeak=4,SideRms=1.7889





Cross correlation image a with $b$, SidePeak $=6$,SideRms $=2.2583$


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## MD and FA probabilities (Pm and Pf) -Plan $\mathrm{C}(\mathrm{a}=0 \times 21 \mathrm{~F} 6, \mathrm{~b}=0 \times \mathrm{C} 9 \mathrm{C} 2)$ at $\mathrm{Eb} / \mathrm{NO}=3 \mathrm{~dB}$ -

From the left figure, the optimum detection threshold is equal to 10 . Then, we can achieve error rate of $1.0 \times 10^{-3}$. If PER performance of payload is higher than $1.0 \times 10^{-3}$ at $\mathrm{Eb} / \mathrm{No}=3 \mathrm{~dB}$, the selected SFD is robust enough not to affect payload performance.

- Correlator values: a

- Correlator values: b

$\operatorname{Pf}(x y)$ denotes that Input signal in correlators is Preamble $+x$ and correlator values are $y$


## PER performance of payload when used FEC

We can achieve bit error rate $(B E R)$ of $1.0 \times 10^{-6}$ at $E b / N o=6 \mathrm{~dB}(\mathrm{CNR}=3 \mathrm{~dB})$ when using the convolutional code $(\mathrm{R}=1 / 2, \mathrm{~K}=4)$. When the payload length is 1500 byte, payload packet error rate (PER) amounts to $1.2 \times 10^{-2}$. Since error rate of start frame delimiting by the SFD $\left(1.3 \times 10^{-3} 1.1 \times 10^{-3}\right.$ or $\left.1.0 \times 10^{-3}\right)$ at $\mathrm{Eb} / \mathrm{No}=3 \mathrm{~dB}(\mathrm{CNR}=3 \mathrm{~dB})$ is better than the PER at the same CNR, the SFD offers sufficient robustness even when FEC is used in payload.

## Non-systematic convolutional code ( $R=1 / 2, K=4$ )



Systematic convolutional code ( $R=1 / 2, K=4$ )


## Proposal on FEC

## Proposed FEC

- Configuration: $\mathrm{r}=1 / 2, \mathrm{~m}=3, \mathrm{n}=2, \mathrm{k}=1, \mathrm{~L}=4, \mathrm{~g} 0=\left\{\begin{array}{llll}1 & 1 & 1 & 1\end{array}\right\}$, $\mathrm{g} 1=\{1101\}$; and feedback connection is set to g 1 as shown in the following figure.
- Free distance is the same as non-systematic convolutional code, which can be calculated from built-in matlab function 'distspec'.
- Tail-bits are inserted according to the shift register values in order to set final state to be 0 .



## BER performance

-Same bit-error performance between sys. conv. [g0,g1]= [17 13,17] and non-sys. conv. [g0,g1]= [17 13]
-Both offers the best performance $\left(\mathrm{Eb} / \mathrm{N} 0=6 \mathrm{~dB}\right.$ at $\left.\mathrm{BER}=10^{\wedge}-6\right)$ in all

Non-systematic convolutional code ( $\mathrm{R}=1 / 2, \mathrm{~K}=4$ )


Systematic convolutional code (Feedback is gO ) $(\mathrm{R}=1 / 2, \mathrm{~K}=4)$


## Reasons to use systematic code

- Merits when envisaged an environment where there are mixed users that have non-FEC mode only and both non-FEC and FEC modes, respectively and some of users are near transmitter and others are far from transmitter
- If used systematic codes, the transmitters have only to send the data encoded by the systematic codes and the receivers can select decoding methods of the received data by using redundancy bit or not. Systematic code based system also permits users to have only non-FEC mode
- If used non-systematic codes, the transmitter needs to decide whether the transmitter uses coding or not in advance and all of users need to have FEC decoding function. Non-Systematic code based system can not permit users who have only non-FEC mode
- Both Systematic and non-systematic convolutional coding have merits and demerits, so the proposal for IEEE802.15.4g may need to combine both of them


## FEC Proposal for IEEE 802.15.4g (1/2)

- Two convolutional codes will be optional FEC

| Mode | R | m | n | k | L | g0 | g1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Systematic | 1/2 | 3 | 2 | 1 | 4 | \{1111\} | \{1011\} |
| Non Systematic | 1/2 | 3 | 2 | 1 | 4 | \{1111\} | \{1011\} |

R : Coding rate
m : number of memory registers
n : number of output bits
k : input bits
L : Constraint length ( $\mathrm{n}+1$ )
g0: Connection vector 0
g1: Connection vector 1

## FEC Proposal for IEEE 802.15.4g (2/2)

## Systematic convolutional code

$\xrightarrow{\text { Output data }} u_{0}$


Convolutional code


